

Isolated, Fixed-Ratio DC-DC Converter

Features & Benefits

- Up to 35 A continuous secondary output current
- Fixed transformation ratio(K) of 1/8
- Up to 797 W/in³ power density
- 97.7% peak efficiency
- Secondary SELV output
- Built-in EMI filtering and In-rush limiting circuit
- Parallel operation for multi-kW arrays
- OV, OC, UV, short circuit and thermal protection
- 4414 package
- High MTBF
- Thermally enhanced VIA[™] package
- PMBus[™] management interface
- Suitable for hot-swap applications

Typical Applications

- 380 V DC Power Distribution
- Information and Communication Technology (ICT) Equipment
- High End Computing Systems
- Automated Test Equipment
- Industrial Systems
- High Density Energy Systems
- Transportation
- Green Buildings and Microgrids

Part Ordering Information

Product Function	Package Length	Package Width	Package Type	Max Primary Input Voltage	Primary Input Range Ratio	Max Secondary Output Voltage	Max Secondary Output Current	Product Grade (Case Temperature)	Option Field
BCM	44	14	x	D1	E	51	35	y	zz
BCM = Bus Converter Module	Length in Inches x 10	Width in Inches x 10	B = Board VIA V = Chassis VIA	Internal Reference				C = -20 to 100°C ^[1] T = -40 to 100°C ^[1]	02 = Chassis/PMBus 06 = Short Pin/PMBus 10 = Long Pin/PMBus

^[1] High Temperature Current Derating may apply; See Figure 1, specified thermal operating area.

Product Ratings	
$V_{PRI} = 400 \text{ V (260 - 410 V)}$	$I_{SEC} = \text{up to } 35 \text{ A}$
$V_{SEC} = 50 \text{ V (32.5 - 51.3 V)}$ (NO LOAD)	$K = 1/8$

Product Description

The HV BCM in a VIA package is a high efficiency Bus Converter, operating from a 260 to 410 V_{DC} primary bus to deliver an isolated 32.5 to 51.3 V_{DC} unregulated, secondary output.

This unique ultra-low profile module incorporates DC-DC conversion, integrated filtering and PMBus[™] commands and controls in a chassis or PCB mount form factor.

The HV BCM offers low noise, fast transient response and industry leading efficiency and power density. A secondary referenced PMBus[™] compatible telemetry and control interface provides access to the HV BCM's internal controller configuration, fault monitoring, and other telemetry functions.

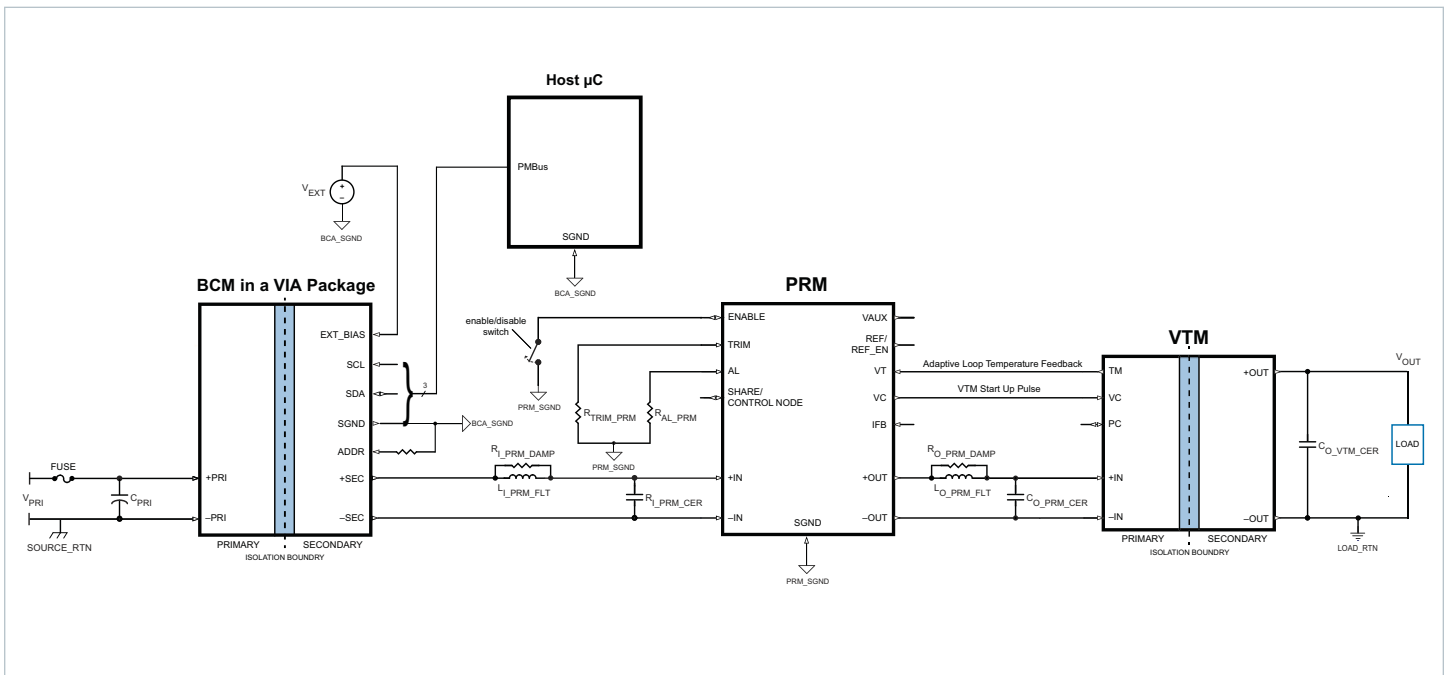
Leveraging the thermal and density benefits of Vicor's VIA packaging technology, the HV BCM module offers flexible thermal management options with very low top and bottom side thermal impedances.

When combined with downstream Vicor DC-DC conversion components and regulators, the BCM allows the Power Design Engineer to employ a simple, low-profile design which will differentiate the end system without compromising on cost or performance metrics.

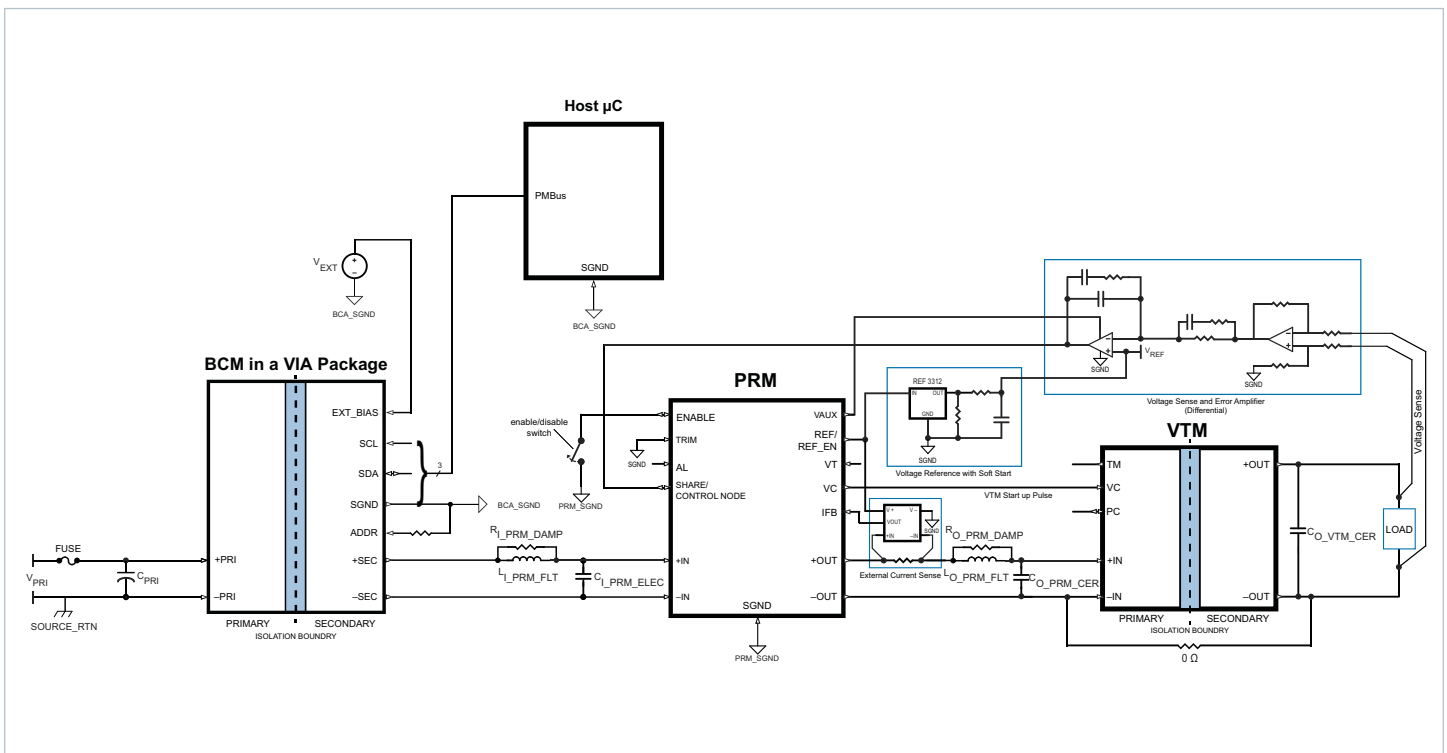


Size:
4.35 x 1.40 x 0.37 in
110.55 x 35.54 x 9.40 mm

Typical Application

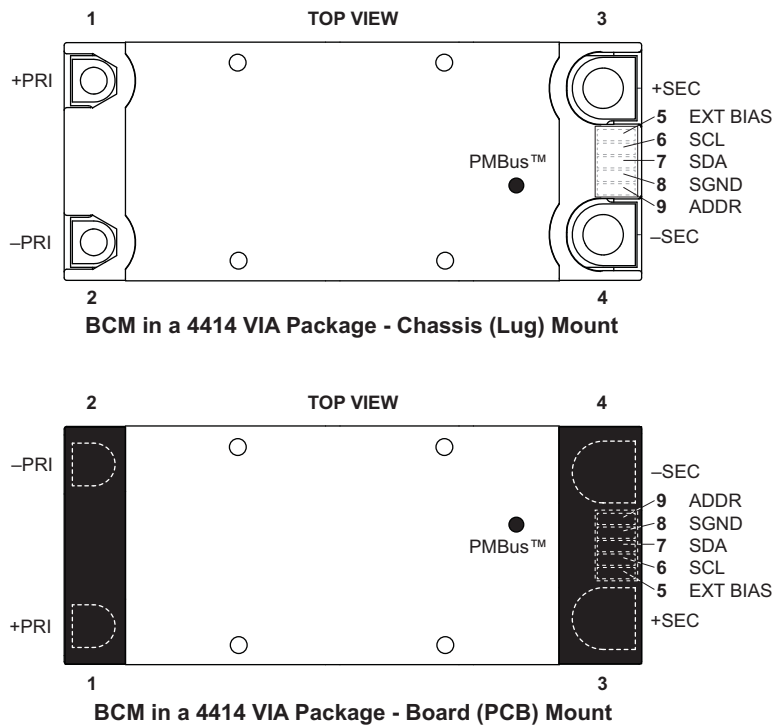


BCM4414xD1E5135yzz + PRM + VTM, Adaptive Loop Configuration



BCM4414xD1E5135yzz + PRM + VTM, Remote Sense Configuration

Pin Configuration



Note: The dot on the VIA housing indicates the location of the signal pin 9.

Pin Descriptions

Pin Number	Signal Name	Type	Function
1	+PRI	PRIMARY POWER	Positive primary transformer power terminal
2	-PRI	PRIMARY POWER RETURN	Negative primary transformer power terminal
3	+SEC	SECONDARY POWER	Positive secondary transformer power terminal
4	-SEC	SECONDARY POWER RETURN	Negative secondary transformer power terminal
5	EXT BIAS	INPUT	5 V supply input
6	SCL	INPUT	I ² C Clock, PMBus Compatible
7	SDA	INPUT/OUTPUT	I ² C Data, PMBus Compatible
8	SGND	SECONDARY SIGNAL RETURN	Signal Ground
9	ADDR	INPUT	Address assignment - Resistor based

Note: All signal pins (5, 6, 7, 8, 9) are referenced to secondary side and isolated from the primary.

Absolute Maximum Ratings

The absolute maximum ratings below are stress ratings only. Operation at or beyond these maximum ratings can cause permanent damage to the device.

Parameter	Comments	Min	Max	Unit
+PRI to -PRI		-1	480	V
PRI_DC or SEC_DC slew rate			N/A	V/ μ s
+SEC to -SEC		-1	60	V
EXT BIAS to SGND		-0.3	10	V
			0.15	A
SCL to SGND		-0.3	5.5	V
SDA to SGND		-0.3	5.5	V
ADDR to SGND		-0.3	3.6	V
Dielectric Withstand*	See note below			
Primary-Case	Basic Insulation	2121		Vdc
Primary-Secondary	Reinforced Insulation (4242 Vdc)	2121		Vdc
Secondary-Case	Functional Insulation	707		Vdc

* Please see Dielectric Withstand section for details regarding test procedure, test values and insulation levels.

Electrical Specifications

Specifications apply over all line and load conditions, unless otherwise noted; **Boldface** specifications apply over the temperature range of $-40^{\circ}\text{C} \leq T_{\text{CASE}} \leq 100^{\circ}\text{C}$ (T-Grade); All other specifications are at $T_{\text{CASE}} = 25^{\circ}\text{C}$ unless otherwise noted.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
General Powetrain PRIMARY to SECONDARY Specification (Forward Direction)						
Primary Input Voltage range, continuous	$V_{\text{PRI_DC}}$		260		410	V
Primary Input Voltage range, transient	$V_{\text{PRI_TRANS}}$		260		410	V
V_{PRI} μ Controller	$V_{\mu\text{C_ACTIVE}}$	$V_{\text{PRI_DC}}$ voltage where μC is initialized, (powertrain inactive)			120	V
PRI to SEC Input Quiescent Current	$I_{\text{PRI_Q}}$	Disabled, $V_{\text{PRI_DC}} = 400\text{ V}$ $T_{\text{CASE}} \leq 100^{\circ}\text{C}$		2	4	mA
PRI to SEC No Load Power Dissipation	$P_{\text{PRI_NL}}$	$V_{\text{PRI_DC}} = 400\text{ V}$, $T_{\text{CASE}} = 25^{\circ}\text{C}$		10.5	17	
		$V_{\text{PRI_DC}} = 400\text{ V}$	6		21	
		$V_{\text{PRI_DC}} = 260\text{ V to }410\text{ V}$, $T_{\text{CASE}} = 25^{\circ}\text{C}$			18	
		$V_{\text{PRI_DC}} = 260\text{ V to }410\text{ V}$			22	
PRI to SEC Inrush Current Peak	$I_{\text{PRI_INR_PK}}$	$V_{\text{PRI_DC}} = 410\text{ V}$, $C_{\text{SEC_EXT}} = 100\ \mu\text{F}$, $R_{\text{LOAD_SEC}} = 25\%$ of full load current $T_{\text{CASE}} \leq 100^{\circ}\text{C}$		6	12	A
DC Primary Input Current	$I_{\text{PRI_IN_DC}}$	At $I_{\text{SEC_OUT_DC}} = 35\text{ A}$, $T_{\text{CASE}} \leq 70^{\circ}\text{C}$			4.5	A
Transformation Ratio	K	Primary to secondary, $K = V_{\text{SEC_DC}} / V_{\text{PRI_DC}}$, at no load		1/8		V/V
Secondary Output Current (continuous)	$I_{\text{SEC_OUT_DC}}$	$T_{\text{CASE}} \leq 70^{\circ}\text{C}$			35	A
Secondary Output Current (pulsed)	$I_{\text{SEC_OUT_PULSE}}$	2 ms pulse, 25% Duty cycle, $I_{\text{SEC_OUT_AVG}} \leq 50\%$ rated $I_{\text{SEC_OUT_DC}}$			40	A
PRI to SEC Efficiency (ambient)	η_{AMB}	$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$	96.5	97.2		%
		$V_{\text{PRI_DC}} = 260\text{ V to }410\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$	95.3			
		$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 17.5\text{ A}$	96.8	97.6		
PRI to SEC Efficiency (hot)	η_{HOT}	$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$, $T_{\text{CASE}} = 70^{\circ}\text{C}$	95.7	96.5		%
PRI to SEC Efficiency (over load range)	$\eta_{20\%}$	$7\text{ A} < I_{\text{SEC_OUT_DC}} < 35\text{ A}$	94.5			%
PRI to SEC Output Resistance	$R_{\text{SEC_COLD}}$	$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$, $T_{\text{CASE}} = -40^{\circ}\text{C}$	18	22	25	m Ω
	$R_{\text{SEC_AMB}}$	$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$	27	29.5	33	
	$R_{\text{SEC_HOT}}$	$V_{\text{PRI_DC}} = 400\text{ V}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$, $T_{\text{CASE}} = 70^{\circ}\text{C}$	32	34.8	37	
Switching Frequency	F_{SW}	Frequency of the Output Voltage Ripple = $2 \times F_{\text{SW}}$	1.05	1.10	1.14	MHz
Secondary Output Voltage Ripple	$V_{\text{SEC_OUT_PP}}$	$C_{\text{SEC_EXT}} = 0\ \mu\text{F}$, $I_{\text{SEC_OUT_DC}} = 35\text{ A}$, $V_{\text{PRI_DC}} = 400\text{ V}$, 20 MHz BW $T_{\text{CASE}} \leq 100^{\circ}\text{C}$		250		mV
					550	

Electrical Specifications (Cont.)

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Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
General Powetrain PRIMARY to SECONDARY Specification (Forward Direction) Cont.						
Effective Primary Capacitance (Internal)	$C_{\text{PRI_INT}}$	Effective Value at 400 $V_{\text{PRI_DC}}$		0.4		μF
Effective Secondary Capacitance (Internal)	$C_{\text{SEC_INT}}$	Effective Value at 50 $V_{\text{SEC_DC}}$		37.6		μF
Effective Secondary Output Capacitance (External)	$C_{\text{SEC_OUT_EXT}}$	Excessive capacitance may drive module into SC protection			100	μF
Effective Secondary Output Capacitance (External)	$C_{\text{SEC_OUT_AEXT}}$	$C_{\text{SEC_OUT_AEXT Max}} = N * 0.5 * C_{\text{SEC_OUT_EXT Max}}$, where N = the number of units in parallel				
Powertrain Protection PRIMARY to SECONDARY (Forward Direction)						
Auto Restart Time	$t_{\text{AUTO_RESTART}}$	Startup into a persistent fault condition. Non-Latching fault detection given $V_{\text{PRI_DC}} > V_{\text{PRI_UVLO+}}$	290		360	ms
Primary Overvoltage Lockout Threshold	$V_{\text{PRI_OVLO+}}$		430	440	450	V
Primary Overvoltage Recovery Threshold	$V_{\text{PRI_OVLO-}}$		420	430	440	V
Primary Overvoltage Lockout Hysteresis	$V_{\text{PRI_OVLO_HYST}}$			10		V
Primary Overvoltage Lockout Response Time	$t_{\text{PRI_OVLO}}$			10		μs
Primary Soft-Start Time	$t_{\text{PRI_SOFT-START}}$	From powertrain active. Fast Current limit protection disabled during Soft-Start		1		ms
Secondary Output Overcurrent Trip Threshold	$I_{\text{SEC_OUT_OCP}}$		37.5	47	59	A
Secondary Output Overcurrent Response Time Constant	$t_{\text{SEC_OUT_OCP}}$	Effective internal RC filter		3.6		ms
Secondary Output Short Circuit Protection Trip Threshold	$I_{\text{SEC_OUT_SCP}}$		52			A
Secondary Output Short Circuit Protection Response Time	$t_{\text{SEC_OUT_SCP}}$			1		μs
Overtemperature Shutdown Threshold	$t_{\text{OTP+}}$	Temperature sensor located inside controller IC (Internal Temperature)	125			$^{\circ}\text{C}$

Electrical Specifications (Cont.)

Specifications apply over all line and load conditions, unless otherwise noted; **Boldface** specifications apply over the temperature range of $-40^{\circ}\text{C} \leq T_{\text{CASE}} \leq 100^{\circ}\text{C}$ (T-Grade); All other specifications are at $T_{\text{CASE}} = 25^{\circ}\text{C}$ unless otherwise noted.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
Powertrain Supervisory Limits PRIMARY to SECONDARY (Forward Direction)						
Primary Overvoltage Lockout Threshold	$V_{\text{PRI_OVLO+}}$		420	436	450	V
Primary Overvoltage Recovery Threshold	$V_{\text{PRI_OVLO-}}$		405	426	440	V
Primary Overvoltage Lockout Hysteresis	$V_{\text{PRI_OVLO_HYST}}$			10		V
Primary Overvoltage Lockout Response Time	$t_{\text{PRI_OVLO}}$			100		μs
Primary Undervoltage Lockout Threshold	$V_{\text{PRI_UVLO-}}$		200	226	250	V
Primary Undervoltage Recovery Threshold	$V_{\text{PRI_UVLO+}}$		225	244	259	V
Primary Undervoltage Lockout Hysteresis	$V_{\text{PRI_UVLO_HYST}}$			15		V
Primary Undervoltage Lockout Response Time	$t_{\text{PRI_UVLO}}$			100		μs
Primary Undervoltage Startup Delay	$t_{\text{PRI_UVLO+_DELAY}}$	From $V_{\text{PRI_DC}} = V_{\text{PRI_UVLO+}}$ to powertrain active, (i.e One time Startup delay form application of $V_{\text{PRI_DC}}$ to $V_{\text{SEC_DC}}$)		20		ms
Secondary Output Overcurrent Trip Threshold	$I_{\text{SEC_OUT_OCP}}$		42.5	45	47.5	A
Secondary Output Overcurrent Response Time Constant	$t_{\text{SEC_OUT_OCP}}$	Effective internal RC filter		2		ms
Overtemperature Shutdown Threshold	$t_{\text{OTP+}}$	Temperature sensor located inside controller IC (Internal Temperature)	125			$^{\circ}\text{C}$
Overtemperature Recovery Threshold	$t_{\text{OTP-}}$	Temperature sensor located inside controller IC (Internal Temperature)	105	110	115	$^{\circ}\text{C}$
Undertemperature Shutdown Threshold	t_{UTP}	Temperature sensor located inside controller IC; Protection not available for M-Grade units.			-45	$^{\circ}\text{C}$
Undertemperature Restart Time	$t_{\text{UTP_RESTART}}$	Startup into a persistent fault condition. Non-Latching fault detection given $V_{\text{PRI_DC}} > V_{\text{PRI_UVLO+}}$		3		s

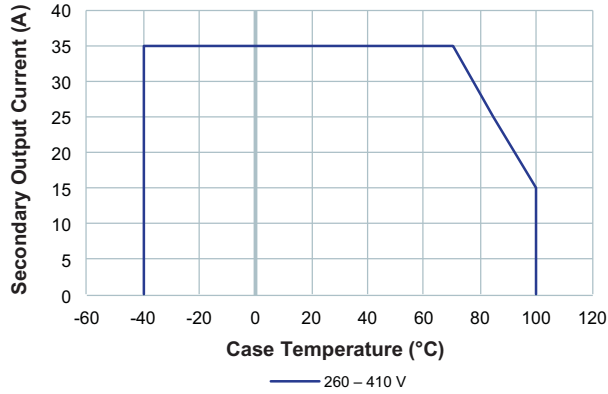


Figure 1 — Specified thermal operating area at high line

1. The BCM in a VIA Package is cooled through bottom case (bottom housing).
2. The thermal rating of the BCM in a VIA Package is based on typical measured device efficiency.
3. The case temperature in the graph is the measured temperature of the bottom housing, such that operating internal junction temperature of the BCM in a VIA Package does not exceed 125°C.

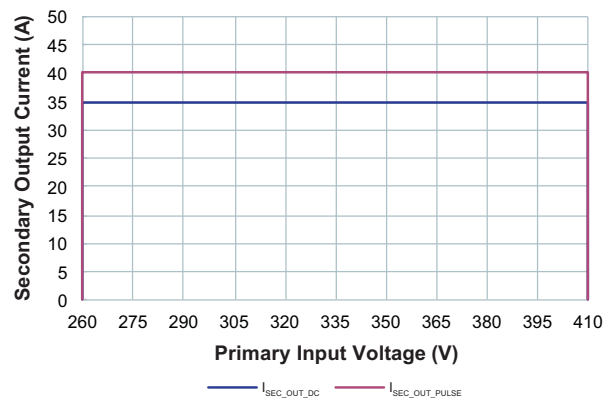
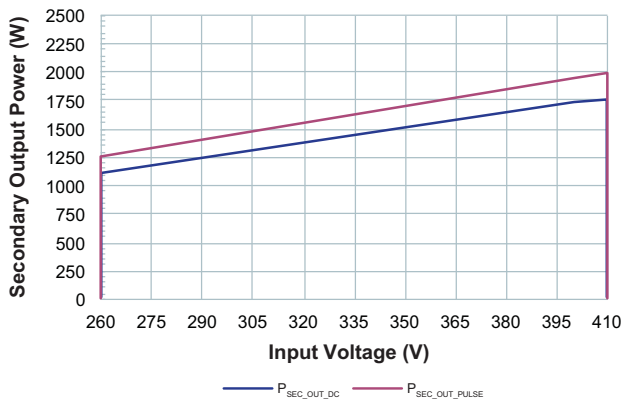


Figure 2 — Specified electrical operating area using rated R_{SEC_HOT}

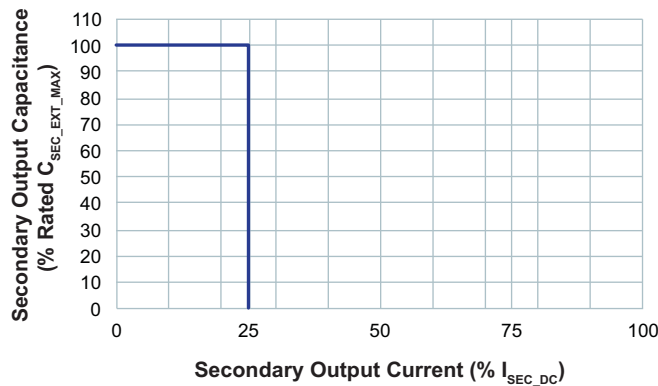


Figure 3 — Specified Primary start-up into load current and external capacitance

PMBus™ Reported Characteristics

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Monitored Telemetry					
• The BCM communication version is not intended to be used without a Digital Supervisor.					
ATTRIBUTE	DIGITAL SUPERVISOR PMBus™ READ COMMAND	ACCURACY (RATED RANGE)	FUNCTIONAL REPORTING RANGE	UPDATE RATE	REPORTED UNITS
Input voltage	(88h) READ_VIN	$\pm 5\%$ (LL - HL)	130 V to 450 V	100 μs	$V_{\text{ACTUAL}} = V_{\text{REPORTED}} \times 10^{-1}$
Input current	(89h) READ_IIN	$\pm 20\%$ (10 - 20% of FL) $\pm 5\%$ (20 - 133% of FL)	-0.85 A to 5.9 A	100 μs	$I_{\text{ACTUAL}} = I_{\text{REPORTED}} \times 10^{-3}$
Output voltage ^[1]	(8Bh) READ_VOUT	$\pm 5\%$ (LL - HL)	16.25 V to 56.25 V	100 μs	$V_{\text{ACTUAL}} = V_{\text{REPORTED}} \times 10^{-1}$
Output current	(8Ch) READ_IOUT	$\pm 20\%$ (10 - 20% of FL) $\pm 5\%$ (20 - 133% of FL)	-7 A to 47.5 A	100 μs	$I_{\text{ACTUAL}} = I_{\text{REPORTED}} \times 10^{-2}$
Output resistance	(D4h) READ_ROUT	$\pm 5\%$ (50 - 100% of FL) at NL $\pm 10\%$ (50 - 100% of FL)(LL - HL)	10 m Ω to 40 m Ω	100 ms	$R_{\text{ACTUAL}} = R_{\text{REPORTED}} \times 10^{-5}$
Temperature ^[2]	(8Dh) READ_TEMPERATURE_1	$\pm 7^{\circ}\text{C}$ (Full Range)	-55°C to 130°C	100 ms	$T_{\text{ACTUAL}} = T_{\text{REPORTED}}$

^[1] Default READ Output Voltage returned when unit is disabled = -300 V.

^[2] Default READ Temperature returned when unit is disabled = -273°C .

Variable Parameter					
• Factory setting of all below Thresholds and Warning limits are 100% of listed protection values. • Variables can be written only when module is disabled either EN pulled low or $V_{\text{IN}} < V_{\text{IN_UVLO}}$. • Module must remain in a disabled mode for 3 ms after any changes to the below variables allowing ample time to commit changes to EEPROM.					
ATTRIBUTE	DIGITAL SUPERVISOR PMBus™ COMMAND ^[3]	CONDITIONS / NOTES	ACCURACY (RATED RANGE)	FUNCTIONAL REPORTING RANGE	DEFAULT VALUE
Input / Output Overvoltage Protection Limit	(55h) VIN_OV_FAULT_LIMIT	$V_{\text{IN_OVLO}}$ is automatically 3% lower than this set point	$\pm 5\%$ (LL - HL)	130 V to 435 V	100%
Input / Output Overvoltage Warning Limit	(57h) VIN_OV_WARN_LIMIT		$\pm 5\%$ (LL - HL)	130 V to 435 V	100%
Input / Output Undervoltage Protection Limit	(D7h) DISABLE_FAULTS	Can only be disabled to a preset default value	$\pm 5\%$ (LL - HL)	130 V to 260 V	100%
Input Overcurrent Protection Limit	(5Bh) IIN_OC_FAULT_LIMIT		$\pm 20\%$ (10 - 20% of FL) $\pm 5\%$ (20 - 133% of FL)	0 to 5.625 A	100%
Input Overcurrent Warning Limit	(5Dh) IIN_OC_WARN_LIMIT		$\pm 20\%$ (10 - 20% of FL) $\pm 5\%$ (20 - 133% of FL)	0 to 5.625 A	100%
Overtemperature Protection Limit	(4Fh) OT_FAULT_LIMIT	Internal Temperature	$\pm 7^{\circ}\text{C}$ (Full Range)	0 to 125°C	100%
Overtemperature Warning Limit	(51h) OT_WARN_LIMIT	Internal Temperature	$\pm 7^{\circ}\text{C}$ (Full Range)	0 to 125°C	100%
Turn on Delay	(60h) TON_DELAY	Additional time delay to the Undervoltage Startup Delay	$\pm 50 \mu\text{s}$	0 to 100 ms	0 ms

^[3] Refer to internal μc datasheet for complete list of supported commands.

Signal Characteristics

Specifications apply over all line and load conditions, unless otherwise noted; **Boldface** specifications apply over the temperature range of $-40^{\circ}\text{C} \leq T_{\text{CASE}} \leq 100^{\circ}\text{C}$ (T-Grade); All other specifications are at $T_{\text{CASE}} = 25^{\circ}\text{C}$ unless otherwise noted.

EXT. BIAS (VDDDB) Pin								
<ul style="list-style-type: none"> 5 V supply power input, required to power the circuitry internal to the BCM in a VIA package for communication signals such as SCL, SDA, ADDR etc Voltage to EXT BIAS pin is needed for PMBus enable and disable control. It is not needed for PMBus monitoring voltage, current, power or temperature. Lower voltage is better. It will help to lower the power dissipation in the internal regulator that is generating 3.3 V voltage for communication circuits. Apply voltage to this pin between 4.5 V and 9 V. The nominal voltage is 5 V. 								
SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	TYP	MAX	UNIT
POWER INPUT	Regular Operation	VDDDB Voltage	V_{VDDDB}		4.5	5	9	V
		VDDDB Current consumption	I_{VDDDB}				50	mA
	Startup	Inrush Current Peak	$I_{\text{VDDDB_INR}}$	V_{VDDDB} Slew Rate = 1 V/ μs		3.5		A
		Turn on time	$t_{\text{VDDDB_ON}}$	From $V_{\text{VDDDB_MIN}}$ to PMBus active		1.5		ms

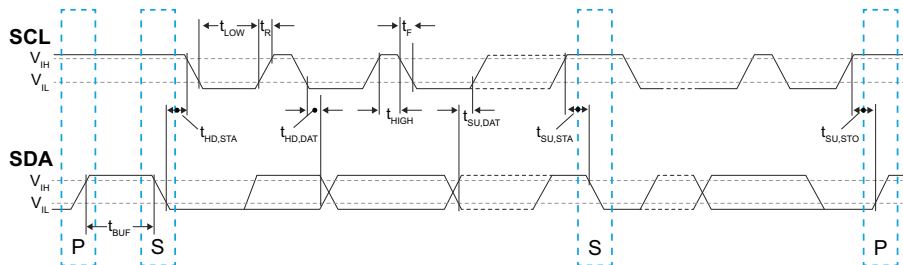
SGND Pin								
<ul style="list-style-type: none"> This pin is power supply return pin for Ext. Bias (VDDDB) pin. All input and output signals (SCL, SDA, ADDR) are referenced to SGND pin. 								

Address (ADDR) Pin								
<ul style="list-style-type: none"> This pin programs only a Fixed and Persistent slave address for BCM in a VIA package. This pin programs the address using a resistor between ADDR pin and signal ground. The address is sampled during startup and is used until power is reset. This pin has 10 kΩ pullup resistor internally between ADDR pin and internal VDD. 16 addresses are available. Relative to nominal value of internal VDD ($V_{\text{VDD_NOM}} = 3.3\text{ V}$), a 206.25 mV range per address. 								
SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	TYP	MAX	UNIT
MULTI-LEVEL INPUT	Regular Operation	ADDR Input Voltage	V_{SADDR}	See address section	0		3.3	V
		ADDR leakage current	I_{SADDR}	Leakage current			1	μA
	Startup	ADDR registration time	t_{SADDR}	From $V_{\text{VDD_IN_MIN}}$		1		ms

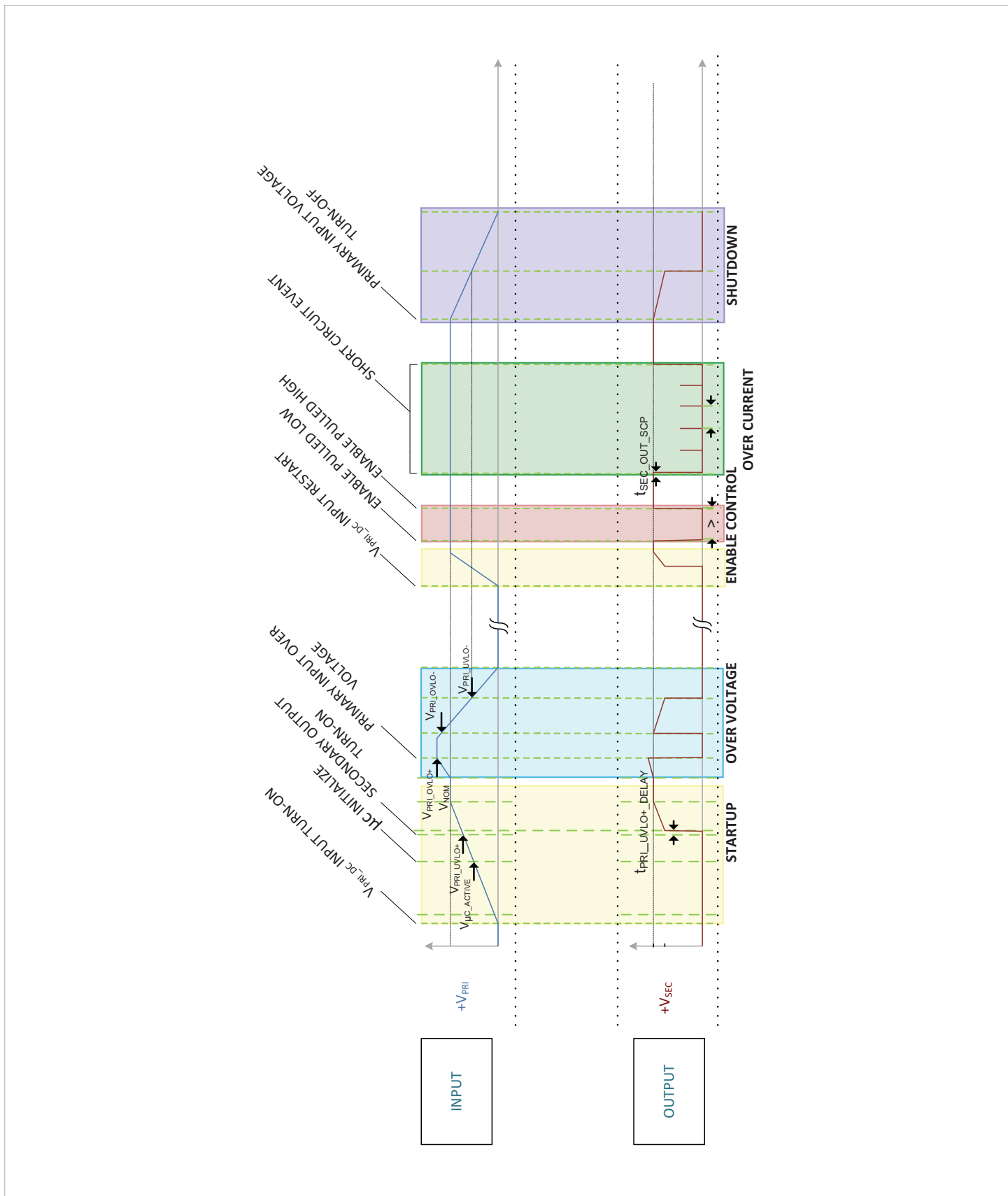
Serial Clock input (SCL) AND Serial Data (SDA) Pins

- High-power SMBus specification and SMBus physical layer compatible. Note that optional SMBALERT# is signal not supported.
- PMBus™ command compatible.
- The internal μC requires the use of a flip-flop to drive SSTOP. See system diagram section for more details.

SIGNAL TYPE	STATE	ATTRIBUTE	SYMBOL	CONDITIONS / NOTES	MIN	TYP	MAX	UNIT
DIGITAL INPUT/OUTPUT	Regular Operation	Electrical Parameters						
		Input Voltage Threshold	V_{IH}	$V_{VDD_IN} = 3.3\text{ V}$	2.1			V
			V_{IL}	$V_{VDD_IN} = 3.3\text{ V}$			0.8	V
		Output Voltage Threshold	V_{OH}	$V_{VDD_IN} = 3.3\text{ V}$	3			V
			V_{OL}	$V_{VDD_IN} = 3.3\text{ V}$			0.4	V
		Leakage current	I_{LEAK_PIN}	Unpowered device			10	μA
		Signal Sink Current	I_{LOAD}	$V_{OL} = 0.4\text{ V}$	4			mA
		Signal Capacitive Load	C_I	Total capacitive load of one device pin			10	pF
		Signal Noise Immunity	V_{NOISE_PP}	10 MHz to 100 MHz	300			mV
		Timing Parameters						
		Operating Frequency	F_{SMB}	Idle state = 0 Hz	10		400	KHz
		Free time between Stop and Start Condition	t_{BUF}		1.3			μs
		Hold time after Start or Repeated Start condition	$t_{HD:STA}$	First clock is generated after this hold time	0.6			μs
		Repeat Start Condition Setup time	$t_{SU:STA}$		0.6			μs
		Stop Condition setup time	$t_{SU:STO}$		0.6			μs
		Data Hold time	$t_{HD:DAT}$		300			ns
		Data Setup time	$t_{SU:DAT}$		100			ns
		Clock low time out	$t_{TIMEOUT}$		25		35	ms
		Clock low period	t_{LOW}		1.3			μs
		Clock high period	t_{HIGH}		0.6		50	μs
Cumulative clock low extend time	$t_{LOW:SEXT}$				25	ms		
Clock or Data Fall time	t_F	Measured from $(V_{IL_MAX} - 0.15)$ to $(V_{IH_MIN} + 0.15)$	20		300	ns		
Clock or Data Rise time	t_R	$0.9 \cdot V_{VDD_IN_MAX}$ to $(V_{IL_MAX} - 0.15)$	20		300	ns		



Timing diagram



Application Characteristics

Product is mounted and temperature controlled via top side cold plate, unless otherwise noted. All data presented in this section are collected data from primary sourced units processing power in forward direction. See associated figures for general trend data.

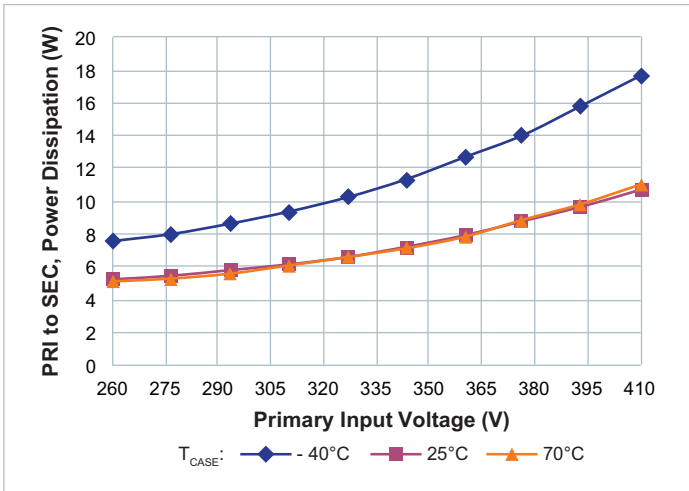


Figure 4 — No load power dissipation vs. V_{PRI_DC}

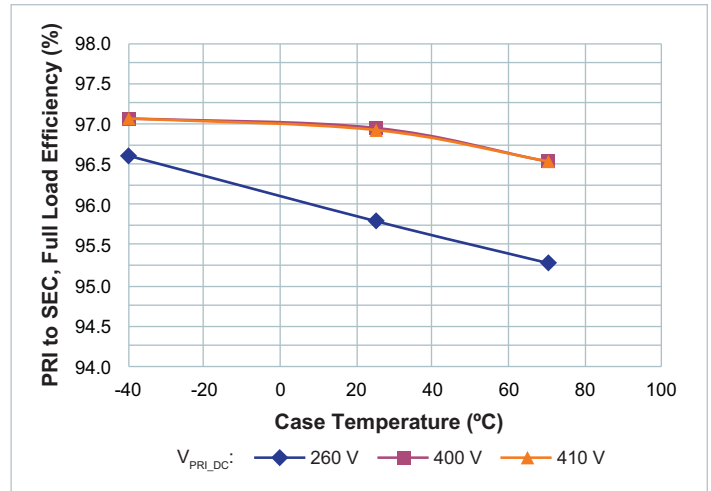


Figure 5 — Full load efficiency vs. temperature; V_{PRI_DC}

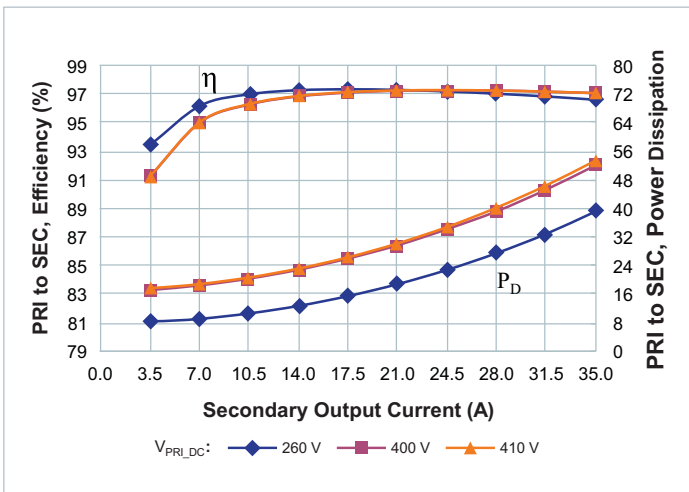


Figure 6 — Efficiency and power dissipation at $T_{CASE} = -40^{\circ}C$

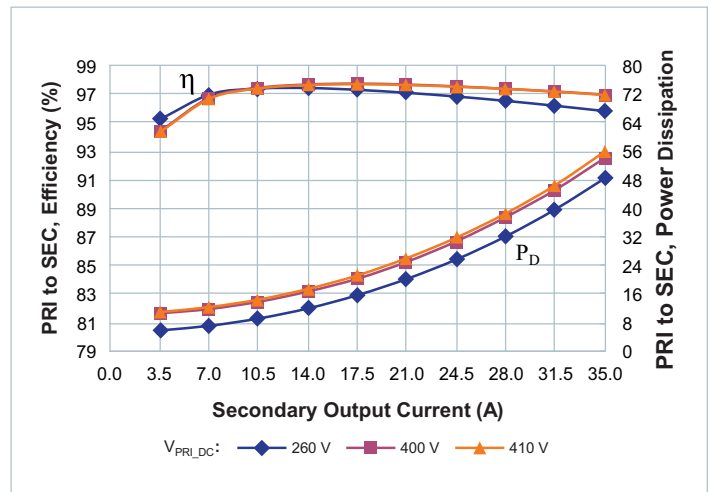


Figure 7 — Efficiency and power dissipation at $T_{CASE} = 25^{\circ}C$

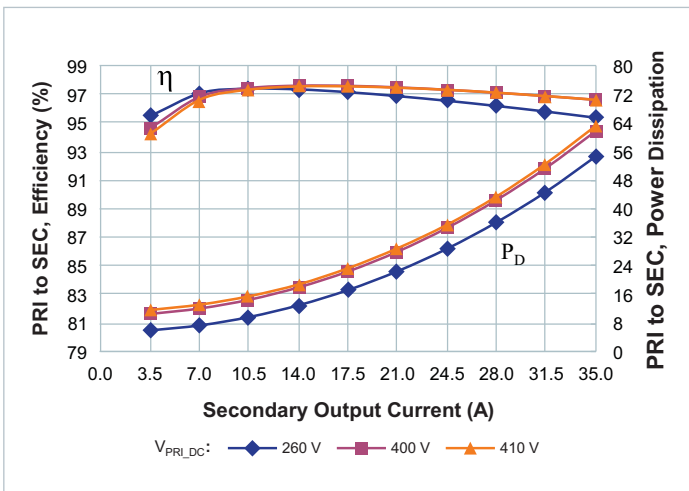


Figure 8 — Efficiency and power dissipation at $T_{CASE} = 70^{\circ}C$

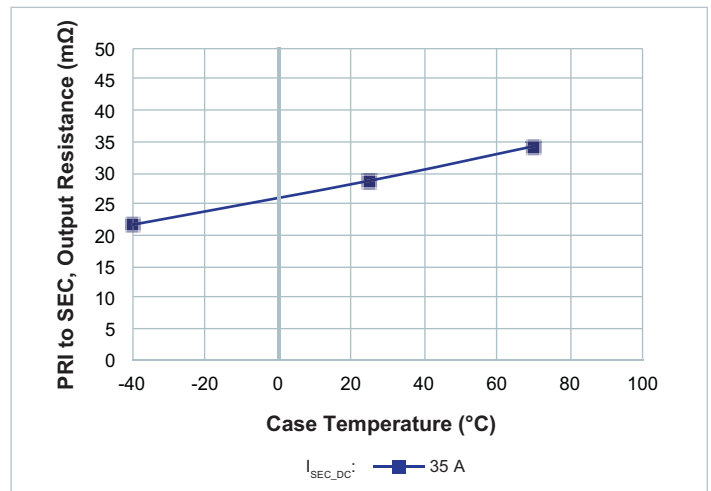


Figure 9 — R_{SEC} vs. temperature; Nominal V_{PRI_DC}
 $I_{SEC_DC} = 35 A$ at $T_{CASE} = 70^{\circ}C$

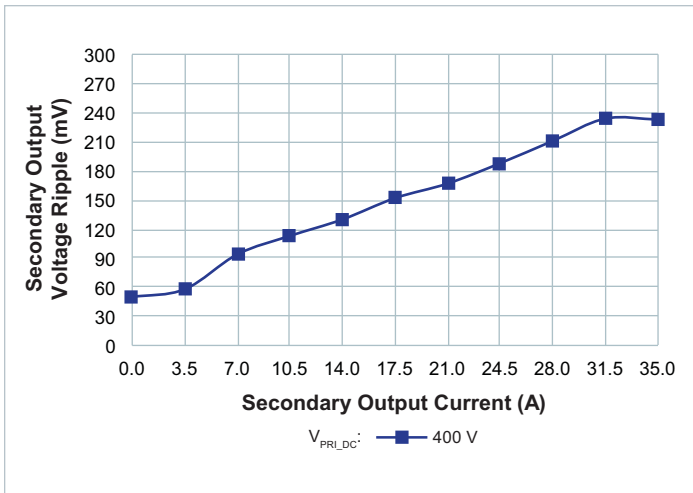


Figure 10 — $V_{SEC_OUT_PP}$ vs. I_{SEC_DC} ; No external $C_{SEC_OUT_EXT}$, Board mounted module, scope setting : 20 MHz analog BW

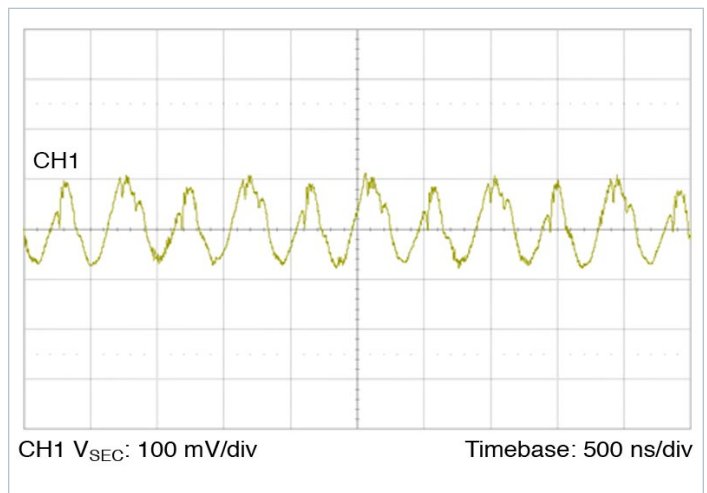


Figure 11 — Full load ripple, $10 \mu F C_{PRI_IN_EXT}$; No external $C_{SEC_OUT_EXT}$, Board mounted module, scope setting : 20 MHz analog BW

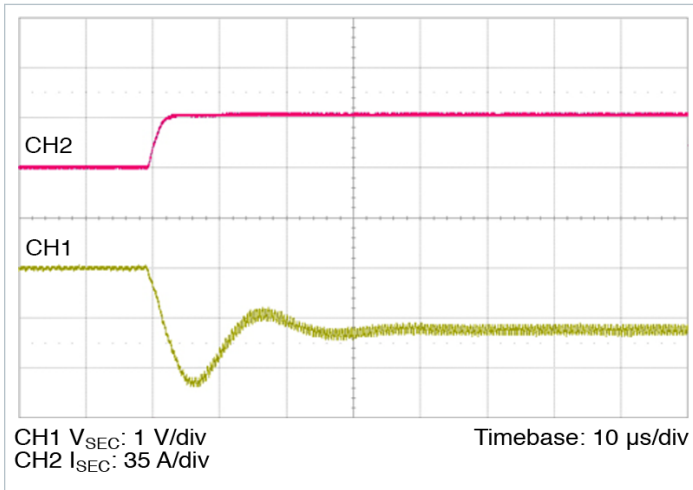


Figure 12 — 0 A–35 A transient response: $C_{PRI_IN_EXT} = 10 \mu F$, no external $C_{SEC_OUT_EXT}$

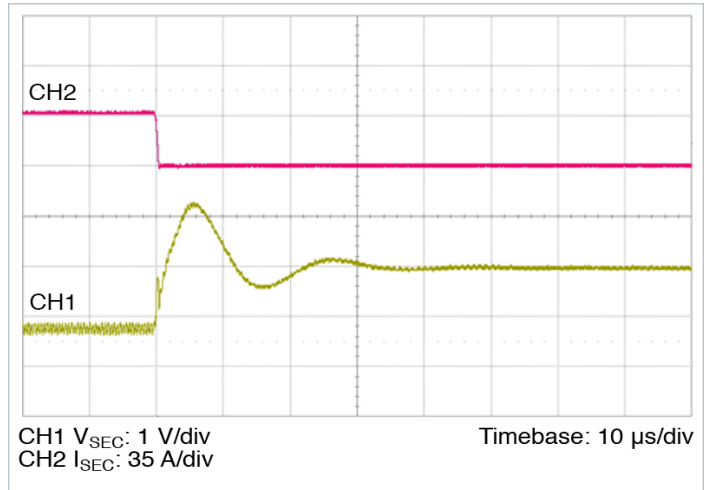


Figure 13 — 35 A–0 A transient response: $C_{PRI_IN_EXT} = 10 \mu F$, no external $C_{SEC_OUT_EXT}$

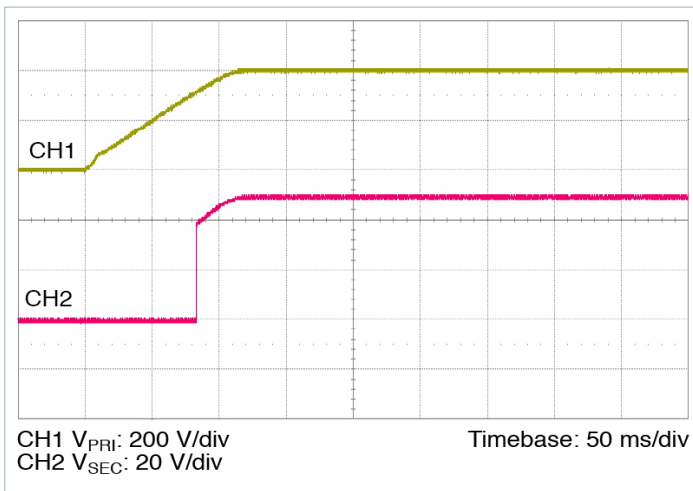


Figure 14 — Start up from application of $V_{PRI_DC} = 400 V$, 25% I_{SEC_DC} 100% $C_{SEC_OUT_EXT}$

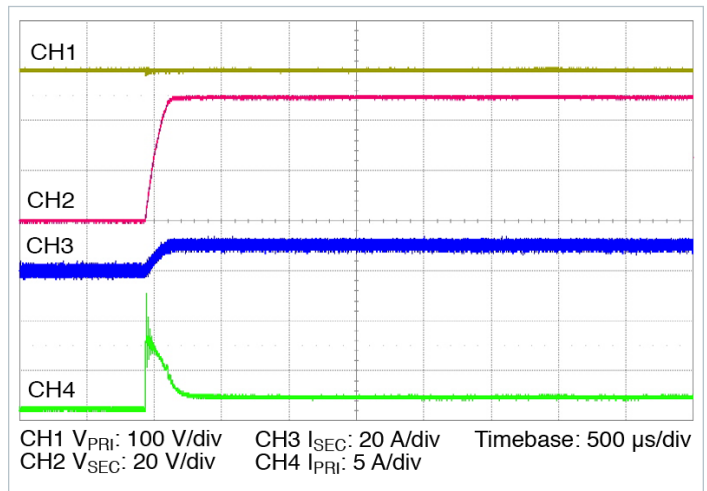


Figure 15 — Start up from application of EN with pre-applied $V_{PRI_DC} = 400 V$, 25% I_{SEC_DC} 100% $C_{SEC_OUT_EXT}$

General Characteristics

Specifications apply over all line and load conditions, unless otherwise noted; **Boldface** specifications apply over the temperature range of $-40^{\circ}\text{C} \leq T_{\text{CASE}} \leq 100^{\circ}\text{C}$ (T-Grade); All other specifications are at $T_{\text{CASE}} = 25^{\circ}\text{C}$ unless otherwise noted.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit	
Mechanical							
Length	L	Lug (Chassis) Mount	110.30 / [4.34]	110.55 / [4.35]	110.80 / [4.36]	mm / [in]	
Length	L	PCB (Board) Mount	112.51 / [4.43]	112.76 / [4.44]	113.01 / [4.45]	mm / [in]	
Width	W		35.29 / [1.39]	35.54 / [1.40]	35.79 / [1.41]	mm / [in]	
Height	H		9.019 / [0.355]	9.40 / [0.37]	9.781 / [0.385]	mm / [in]	
Volume	Vol	Without heatsink		36.93 / [2.25]		cm ³ / [in ³]	
Weight	W			140.5 / [4.96]		g / [oz]	
Pin Material		C145 copper, 1/2 hard					
Underplate		Low stress ductile Nickel	50		100	μin	
Pin Finish		Palladium	0.8		6	μin	
		Soft Gold	0.12		2		
Thermal							
Operating junction temperature	T _{INTERNAL}	BCM4414xD1E5135yzz (T-Grade)	-40		125	°C	
		BCM4414xD1E5135yzz (C-Grade)	-20		125		
Operating case temperature	T _{CASE}	BCM4414xD1E5135yzz (T-Grade), derating applied, see safe thermal operating area	-40		100		
		BCM4414xD1E5135yzz (C-Grade), derating applied, see safe thermal operating area	-20		100		
Thermal resistance top side	R _{JC-TOP}	Estimated thermal resistance to maximum temperature internal component from isothermal top		1.24			°C/W
Thermal Resistance Coupling between top case and bottom case	R _{HOU}	Estimated thermal resistance of thermal coupling between the top and bottom case surfaces		0.63			°C/W
Thermal resistance bottom side	R _{JC-BOT}	Estimated thermal resistance to maximum temperature internal component from isothermal bottom		1.41		°C/W	
Thermal capacity				54		Ws/°C	
Assembly							
Storage Temperature	T _{ST}	BCM4414xD1E5135yzz (T-Grade)	-40		125	°C	
		BCM4414xD1E5135yzz (C-Grade)	-40		125	°C	
ESD Withstand	ESD _{HBM}	Human Body Model, "ESDA / JEDEC JDS-001-2012" Class I-C (1kV to < 2 kV)	1000				
	ESD _{CDM}	Charge Device Model, "JESD 22-C 101-E" Class II (200V to < 500V)	200				

General Characteristics (Cont.)

Specifications apply over all line and load conditions, unless otherwise noted; **Boldface** specifications apply over the temperature range of $-40^{\circ}\text{C} \leq T_{\text{CASE}} \leq 100^{\circ}\text{C}$ (T-Grade); All other specifications are at $T_{\text{CASE}} = 25^{\circ}\text{C}$ unless otherwise noted.

Attribute	Symbol	Conditions / Notes	Min	Typ	Max	Unit
Safety						
Isolation capacitance	C _{IN_OUT}	Unpowered unit	620	780	940	pF
Isolation resistance	R _{IN_OUT}	At 500 Vdc	10			MΩ
MTBF		MIL-HDBK-217Plus Parts Count - 25°C Ground Benign, Stationary, Indoors / Computer		3.53		MHrs
		Telcordia Issue 2 - Method I Case III; 25°C Ground Benign, Controlled		3.90		MHrs
Agency approvals / standards		cTÜVus "EN 60950-1 "				
		cURus "UL 60950-1 "				
		CE Marked for Low Voltage Directive and RoHS Recast Directive, as applicable				

BCM in a VIA Package

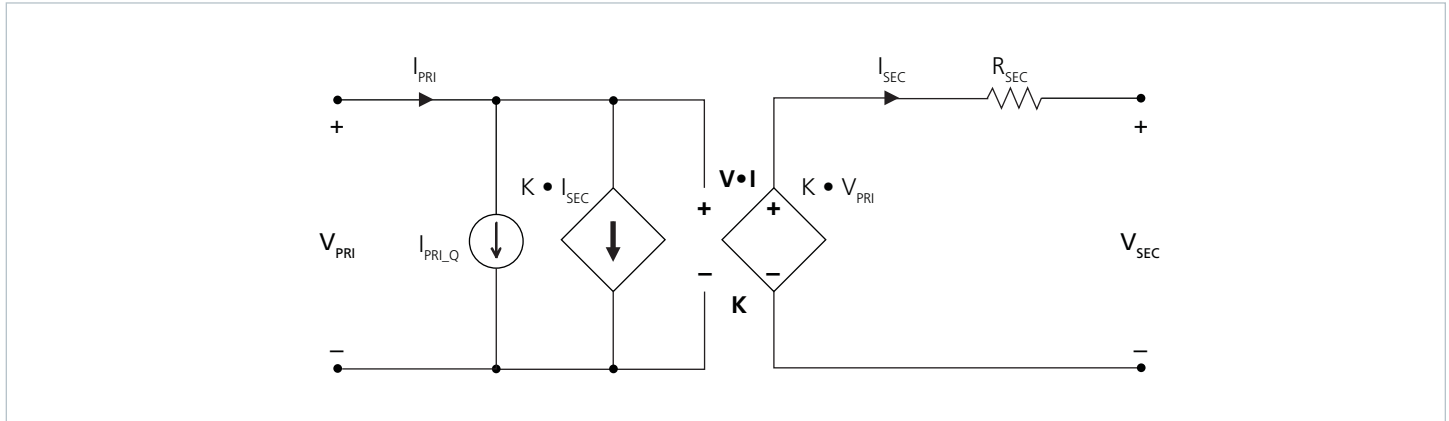


Figure 16 – BCM DC model

The BCM in a VIA package uses a high frequency resonant tank to move energy from Primary to secondary and vice versa. (The resonant tank is formed by Cr and leakage inductance Lr in the power transformer windings as shown in the BCM module Block Diagram). The resonant LC tank, operated at high frequency, is amplitude modulated as a function of input voltage and output current. A small amount of capacitance embedded in the primary and secondary stages of the module is sufficient for full functionality and is key to achieving high power density.

The BCM4414xD1E5135yzz can be simplified into the preceding model.

At no load:

$$V_{SEC} = V_{PRI} \cdot K \tag{1}$$

K represents the “turns ratio” of the BCM. Rearranging Eq (1):

$$K = \frac{V_{SEC}}{V_{PRI}} \tag{2}$$

In the presence of load, V_{SEC} is represented by:

$$V_{SEC} = V_{PRI} \cdot K - I_{SEC} \cdot R_{SEC} \tag{3}$$

and I_{SEC} is represented by:

$$I_{SEC} = \frac{I_{PRI} - I_{PRI_Q}}{K} \tag{4}$$

R_{SEC} represents the impedance of the BCM, and is a function of the R_{DSON} of the input and output MOSFETs, PC board resistance of input and output boards and the winding resistance of the power transformer. I_Q represents the quiescent current of the BCM control, gate drive circuitry, and core losses.

The use of DC voltage transformation provides additional interesting attributes. Assuming that R_{SEC} = 0 Ω and I_{PRI_Q} = 0 A, Eq. (3) now becomes Eq. (1) and is essentially load independent, resistor R is now placed in series with V_{PRI}.

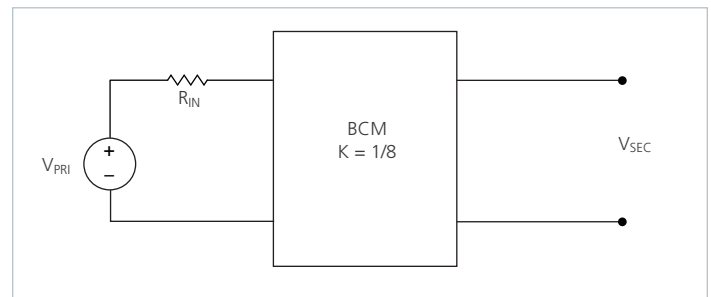


Figure 17 — K = 1/8 BCM with series input resistor

The relationship between V_{PRI} and V_{SEC} becomes:

$$V_{SEC} = (V_{PRI} - I_{PRI} \cdot R_{IN}) \cdot K \tag{5}$$

Substituting the simplified version of Eq. (4) (I_{PRI_Q} is assumed = 0 A) into Eq. (5) yields:

$$V_{SEC} = V_{PRI} \cdot K - I_{SEC} \cdot R_{IN} \cdot K^2 \tag{6}$$

This is similar in form to Eq. (3), where R_{SEC} is used to represent the characteristic impedance of the BCM™. However, in this case a real R on the primary side of the BCM is effectively scaled by K^2 with respect to the secondary.

Assuming that $R = 1 \Omega$, the effective R as seen from the secondary side is 15.6 mΩ, with $K = 1/8$.

A similar exercise should be performed with the addition of a capacitor or shunt impedance at the primary input to the BCM. A switch in series with V_{PRI} is added to the circuit. This is depicted in Figure 18.

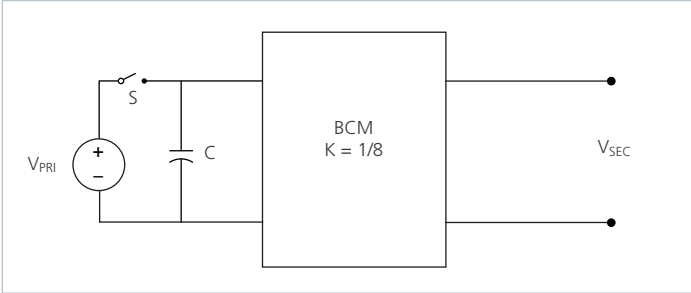


Figure 18 — BCM with input capacitor

A change in V_{PRI} with the switch closed would result in a change in capacitor current according to the following equation:

$$I_C(t) = C \frac{dV_{PRI}}{dt} \quad (7)$$

Assume that with the capacitor charged to V_{PRI} , the switch is opened and the capacitor is discharged through the idealized BCM. In this case,

$$I_C = I_{SEC} \cdot K \quad (8)$$

substituting Eq. (1) and (8) into Eq. (7) reveals:

$$I_{SEC} = \frac{C}{K^2} \cdot \frac{dI_{SEC}}{dt} \quad (9)$$

The equation in terms of the output has yielded a K^2 scaling factor for C, specified in the denominator of the equation.

A K factor less than unity results in an effectively larger capacitance on the secondary output when expressed in terms of the input. With a $K = 1/8$ as shown in Figure 18, $C = 1 \mu F$ would appear as $C = 64 \mu F$ when viewed from the secondary.

Low impedance is a key requirement for powering a high-current, low-voltage load efficiently. A switching regulation stage should have minimal impedance while simultaneously providing appropriate filtering for any switched current. The use of a BCM between the regulation stage and the point of load provides a dual benefit of scaling down series impedance leading back to the source and scaling up shunt capacitance or energy storage as a function of its K factor squared. However, the benefits are not useful if the series impedance of the BCM is too high. The impedance of the BCM must be low, i.e. well beyond the crossover frequency of the system.

A solution for keeping the impedance of the BCM low involves switching at a high frequency. This enables small magnetic components because magnetizing currents remain low. Small magnetics mean small path lengths for turns. Use of low loss core material at high frequencies also reduces core losses.

The two main terms of power loss in the BCM module are:

- No load power dissipation (P_{PRI_NL}): defined as the power used to power up the module with an enabled powertrain at no load.
- Resistive loss (R_{SEC}): refers to the power loss across the BCM® module modeled as pure resistive impedance.

$$P_{DISSIPATED} = P_{PRI_NL} + P_{R_{SEC}} \quad (10)$$

Therefore,

$$P_{SEC_OUT} = P_{PRI_IN} - P_{DISSIPATED} = P_{PRI_IN} - P_{PRI_NL} - P_{R_{SEC}} \quad (11)$$

The above relations can be combined to calculate the overall module efficiency:

$$\eta = \frac{P_{SEC_OUT}}{P_{PRI_IN}} = \frac{P_{PRI_IN} - P_{PRI_NL} - P_{R_{SEC}}}{P_{PRI_IN}} \quad (12)$$

$$= \frac{V_{PRI} \cdot I_{PRI} - P_{PRI_NL} - (I_{SEC})^2 \cdot R_{SEC}}{V_{PRI} \cdot I_{PRI}}$$

$$= 1 - \left(\frac{P_{PRI_NL} + (I_{SEC})^2 \cdot R_{SEC}}{V_{PRI} \cdot I_{PRI}} \right)$$

Thermal Considerations

The VIA™ package provides effective conduction cooling from either of the two module surfaces. Heat may be removed from the top surface, the bottom surface or both. The extent to which these two surfaces are cooled is a key component for determining the maximum power that can be processed by a VIA, as can be seen from specified thermal operating area in Figure 1. Since the VIA has a maximum internal temperature rating, it is necessary to estimate this internal temperature based on a system-level thermal solution. To this purpose, it is helpful to simplify the thermal solution into a roughly equivalent circuit where power dissipation is modeled as a current source, isothermal surface temperatures are represented as voltage sources and the thermal resistances are represented as resistors. Figure 19 shows the “thermal circuit” for the VIA module.

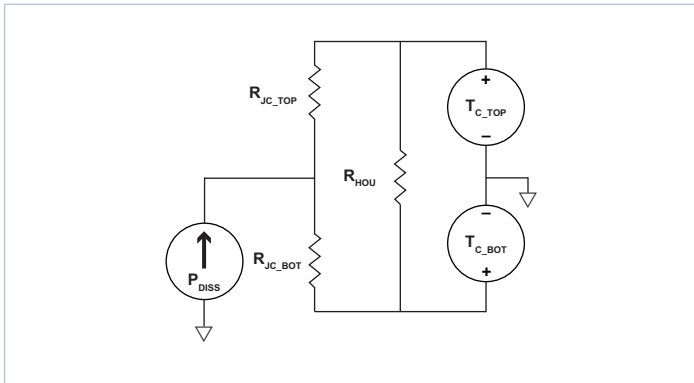


Figure 19 – Double sided cooling VIA thermal model

In this case, the internal power dissipation is P_{DISS} , R_{JC_TOP} and R_{JC_BOT} are thermal resistance characteristics of the VIA module and the top and bottom surface temperatures are represented as T_{C_TOP} and T_{C_BOT} . It is interesting to notice that the package itself provides a high degree of thermal coupling between the top and bottom case surfaces (represented in the model by the resistor R_{HOU}). This feature enables two main options regarding thermal designs:

- Single side cooling: the model of Figure 19 can be simplified by calculating the parallel resistor network and using one simple thermal resistance number and the internal power dissipation curves; an example for bottom side cooling only is shown in Figure 20.

In this case, R_{JC} can be derived as following:

$$R_{JC} = \frac{(R_{JC_TOP} + R_{HOU}) \cdot R_{JC_BOT}}{R_{JC_TOP} + R_{HOU} + R_{JC_BOT}} \quad (14)$$

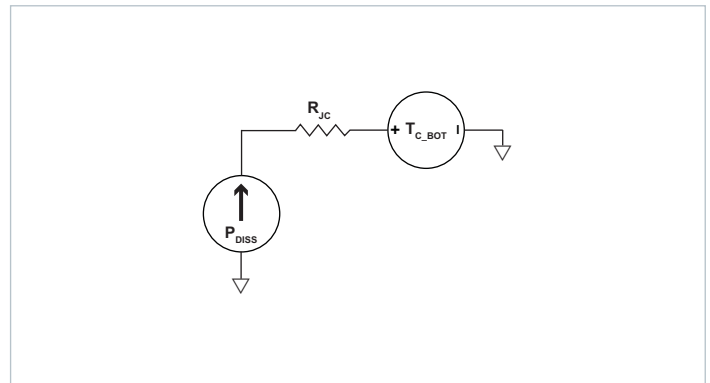


Figure 20 – Single-sided cooling VIA thermal model

- Double side cooling: while this option might bring limited advantage to the module internal components (given the surface-to-surface coupling provided), it might be appealing in cases where the external thermal system requires allocating power to two different elements, like for example heatsinks with independent airflows or a combination of chassis/air cooling.

Current Sharing

The performance of the BCM in a VIA package is based on efficient transfer of energy through a transformer without the need of closed loop control. For this reason, the transfer characteristic can be approximated by an ideal transformer with a positive temperature coefficient series resistance.

This type of characteristic is close to the impedance characteristic of a DC power distribution system both in dynamic (AC) behavior and for steady state (DC) operation.

When multiple BCM modules of a given part number are connected in an array they will inherently share the load current according to the equivalent impedance divider that the system implements from the power source to the point of load.

Some general recommendations to achieve matched array impedances include:

- Dedicate common copper planes/wires within the PCB/Chassis to deliver and return the current to the VIA modules.
- Provide as symmetric a PCB/Wiring layout as possible among VIA™ modules

For further details see [AN:016 Using BCM Bus Converters in High Power Arrays.](#)

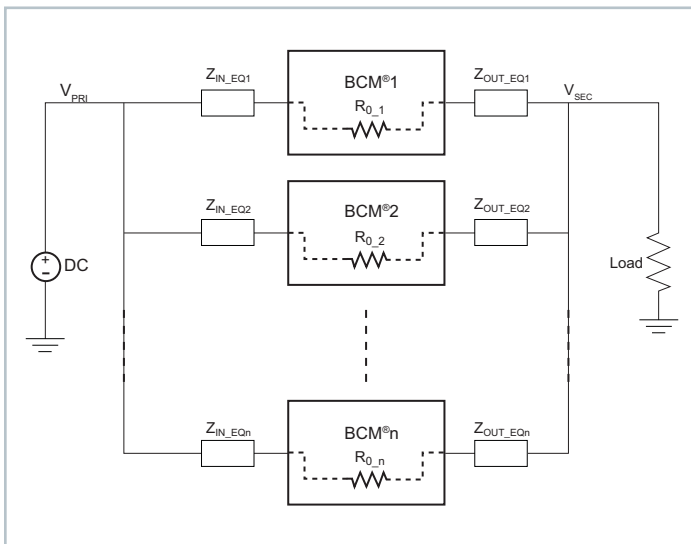


Figure 21 — BCM module array

Fuse Selection

In order to provide flexibility in configuring power systems, BCM in a VIA package modules are not internally fused. Input line fusing of BCM in a VIA package products is recommended at system level to provide thermal protection in case of catastrophic failure.

The fuse shall be selected by closely matching system requirements with the following characteristics:

- Current rating
(usually greater than maximum current of BCM module)
- Maximum voltage rating
(usually greater than the maximum possible input voltage)
- Ambient temperature
- Nominal melting I^2t
- Recommend fuse: 10 A Littelfuse 505 Series or 10 A Littelfuse 487 Series

Reverse Operation

BCM modules are capable of reverse power operation. Once the unit is started, energy will be transferred from secondary back to the primary whenever the secondary voltage exceeds $V_{PRI} \cdot K$. The module will continue operation in this fashion for as long as no faults occur.

The BCM4414xD1E5135yzz has not been qualified for continuous operation in a reverse power condition. Furthermore fault protections which help protect the module in forward operation will not fully protect the module in reverse operation.

Transient operation in reverse is expected in cases where there is significant energy storage on the output and transient voltages appear on the input.

Dielectric Withstand

The chassis of the HV BCM in a VIA Package is required to be connected to Protective Earth when installed in the end application and must satisfy the requirements of IEC 60950-1 for Class I products.

The HV BCM in a VIA Package contains an internal safety approved isolating component (VI ChiP) that provides the Reinforced Insulation from Primary Input to Secondary Output. The isolating component is individually tested for Reinforced Insulation from Primary Input to Secondary Output at 4242 Vdc prior to the final assembly of the VIA™.

When the VIA™ assembly is complete the Reinforced Insulation can only be tested at Basic Insulation values as specified in the electric strength Test Procedure noted in clause 5.2.2 of IEC 60950-1.

Test Procedure Note from IEC 60950-1

“For equipment incorporating both REINFORCED INSULATION and lower grades of insulation, care is taken that the voltage applied to the REINFORCED INSULATION does not overstress BASIC INSULATION or SUPPLEMENTARY INSULATION.”

Summary

The final VIA assembly provides basic insulation from Primary Input to case, reinforced insulation from Primary Input to Secondary Output and functional insulation from Secondary Output to case. Both sides of the housing are required to be connected to Protective Earth to satisfy safety and EMI requirements. Protective earthing can be accomplished through dedicated wiring harness (example: ring terminal clamped by mounting screw) or surface contact (example: pressure contact on bare conductive chassis or PCB copper layer with no solder mask).

The case is required to be connected to protective earth in the final installation. The ground connection of the top case has to be 3 orders of magnitude more resistive than the current return connection to the bottom case. The bottom case connection is expected to be 100 $\mu\Omega$ (or less). Therefore, the grounding connection of the top case should be greater than 100 m Ω . The construction of the VIA can be summarized by describing it as a “Class II” component installed in a “Class I” subassembly. The insulation from Primary Input to Secondary Output can only be tested at basic insulation values on the completely assembled VIA product.

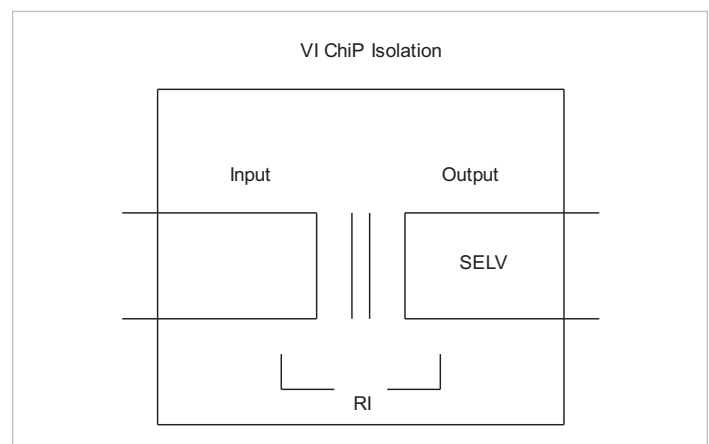


Figure 22 – VI Chip before final assembly in the VIA

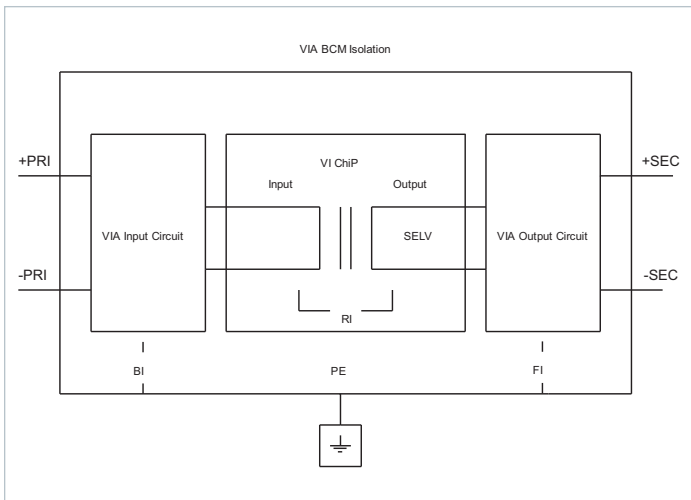


Figure 23 – BCM in a VIA package after final assembly

Input and Output Filtering

The HV BCM in a VIA Package has built-in single stage EMI filtering with hot-swap circuitry primary input side. Typical test set-up block diagram for conducted emissions is shown in Figure 24. Hot-swap circuitry provides inrush current limiting through the MOSFET.

Further, along with internal ceramic input and output capacitance, it reduces the input and output voltage ripple. External output filtering can be added as needed. Ceramic capacitance can be used as an output bypass for this purpose. Moreover, along with hot-swap circuitry, it protects the adapter from overvoltage transients imposed by a system that would exceed maximum ratings and induce stresses. Adapter input and output voltage ranges shall not be exceeded. An internal overvoltage function prevents operation outside of the normal operating input range. Even when disabled, the adapter is exposed to the applied voltage and adapter must withstand it.

Given the wide bandwidth of the adapter, the source response is generally the limiting factor in the overall system response. Anomalies in the response of the source will appear at the output of the module multiplied by its K factor.

Total load capacitance at the output of the adapter shall not exceed the specified maximum for correct operation of it in startup and steady state conditions. Owing to the wide bandwidth and low output impedance of the adapter, low frequency bypass capacitance and significant energy storage may be more densely and efficiently provided by adding capacitance at the input of the adapter. At frequencies less than 500 KHz, the adapter appears as an impedance of R_{OUT} between the source and load.

Within this frequency range, capacitance at the input appears as effective capacitance on the output per the relationship defined in equation (15).

This enable a reduction in the size and number of capacitors used in a typical system.

$$C_{OUT} = \frac{C_{IN}}{K^2} \quad (15)$$

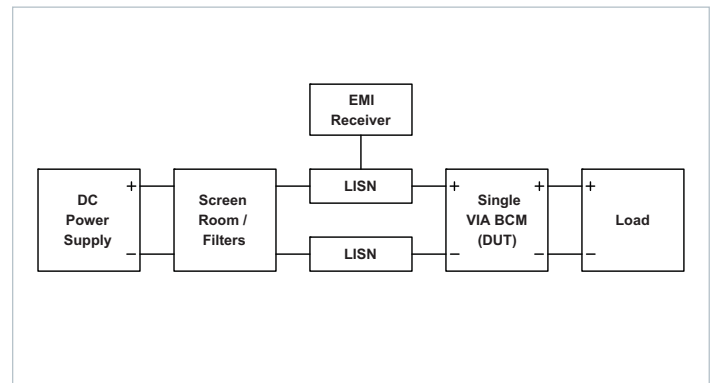


Figure 24 – Typical test set-up block diagram for Conducted Emissions

Hot Swap

Many applications use a power architecture based on a 380 Vdc distribution bus. This supply level is emerging as a new standard and efficient means for distributing power through boards, racks and chassis mounted Telecom and Datacom system. The interconnect between the different modules is accomplished with a backplane and motherboard. Power is commonly provided to the various module slots via a 380 Vdc distribution bus.

Removing the faulty module from the rack is relatively easy, provided the remaining power modules can support the step increase in load. Plugging in the replacement module has more potential for problems, as it will present an uncharged capacitor load and draw a large inrush current. This could cause a momentary, but unacceptable interruption or sag to the backplane power bus if not limited. The problem can also arise if ordinary power module connectors are used, since the connector pins will engage and disengage in a random and unpredictable sequence during insertion and removal.

Hot swap or hot plug is the highly desirable feature in many applications, but it also creates several issues that must be addressed in the system design. A number of related phenomena occur with a live insertion and removal event, including bouncing, arcing between input connector pins, larger voltage and current transients. Hot swap circuitry in the converter modules protects the module and the rest of the system from the problem associated with live insertion.

To meet the maintenance, reconfiguration, redundancy and system upgrade, this new HV BCM module is being designed to address the function of hot-swapping at the 380 Vdc distribution bus. This new module provides a high level of integration for DC to DC converters in 380 Vdc distributions, saving the system designer design time and critical board space. Hot swap circuitry as shown in Figure 29 uses an active MOSFET switching device in the input line. During insertion, the MOSFET is driven into a resistive state to limit the inrush current, and then when the inserted module's input capacitor has charged, the MOSFET becomes fully conductive to avoid the voltage drop losses. PERFORMANCE verification is further illustrated through scope plots of circuit's response to various live insertion events.

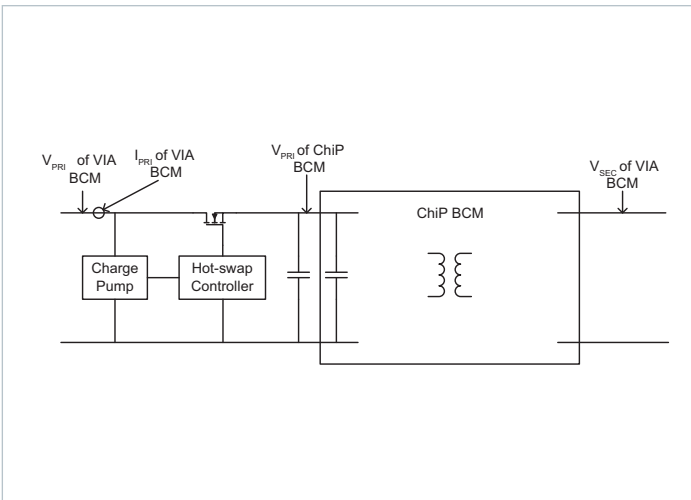


Figure 25 – High-Level Diagram for 384 Vdc Hot Swap with ChiP BCM DC-DC converter

Hot Swap Test – Scope Pictures

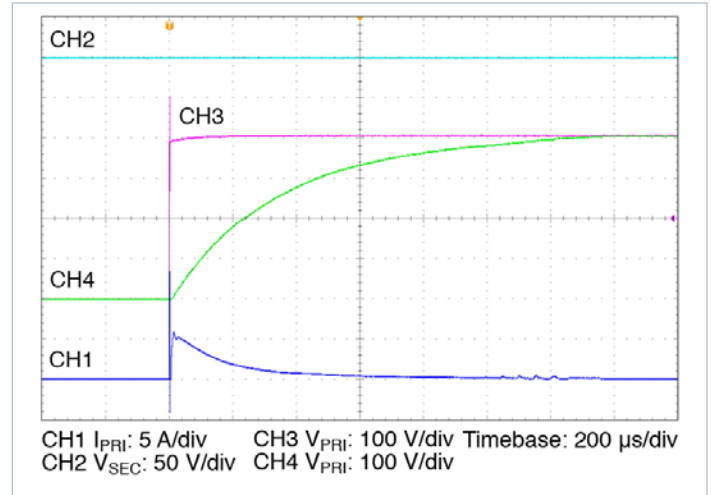


Figure 27 – Hot swap start-up

Overall, the objective is always remains the same in hot swap applications; to give system designer the opportunity to build hot swap capability into redundant power module arrays. This allows telecoms and other mission critical applications to continue without interruption even through failure and replacement of one or possibly more of their power modules.

Hot Swap Test – Test circuit and Procedure

- Two HV BCMs in parallel with mercury relay#1 open
- Close mercury relay#1 and measure inrush current going into #2 HV BCM

- Ch1: I_{PRI} of HV BCM#2
- Ch2: V_{SEC} of HV BCM#2
- Ch3: V_{PRI} of HV BCM#2 shows the fast input voltage transient at the input terminal of HV BCM#2
- Ch4: V_{PRI} of ChiP BCM#2 shows the soft start charging the input capacitor as shown, time constant depends upon the gate signal.

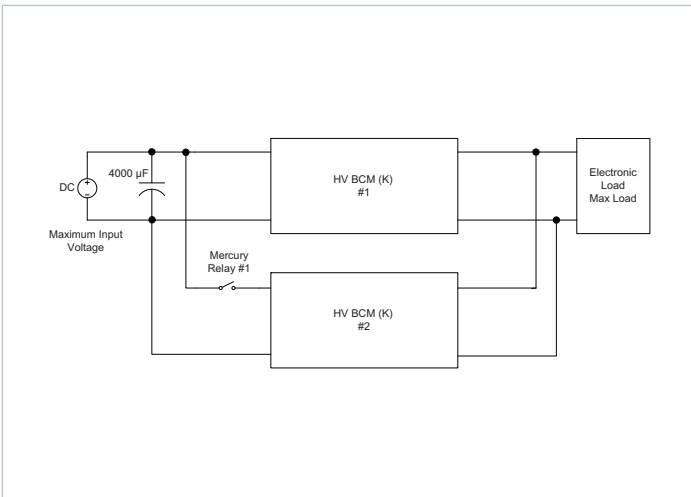


Figure 26 – Test Circuit

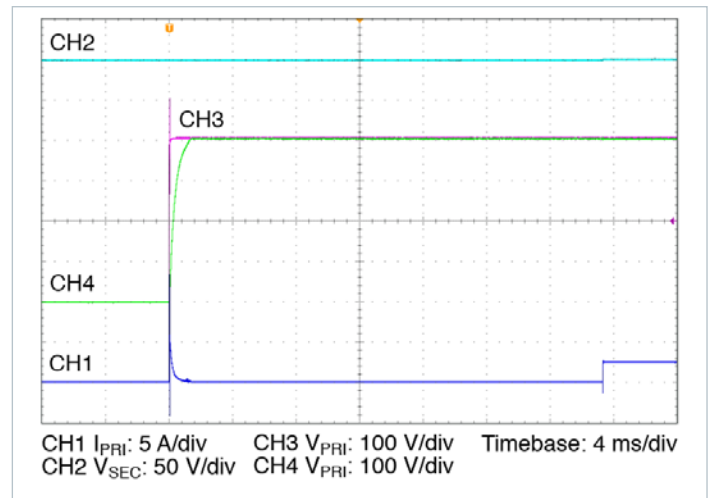
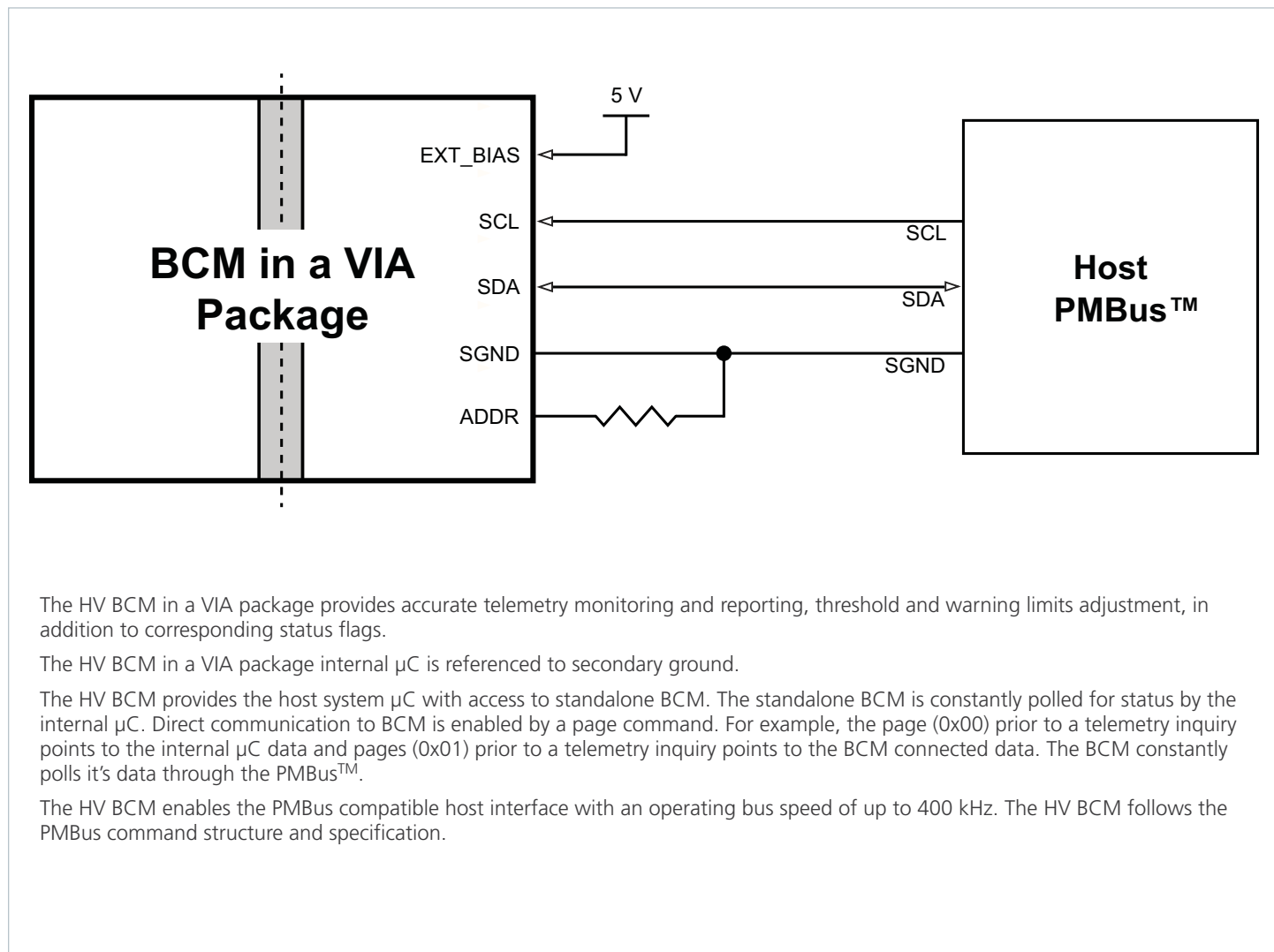


Figure 28 – Same as figure 30 but at a bigger time scale shows the appearance of the HV BCM#2 output

System Diagram for PMBus™ Interface



The HV BCM in a VIA package provides accurate telemetry monitoring and reporting, threshold and warning limits adjustment, in addition to corresponding status flags.

The HV BCM in a VIA package internal μC is referenced to secondary ground.

The HV BCM provides the host system μC with access to standalone BCM. The standalone BCM is constantly polled for status by the internal μC . Direct communication to BCM is enabled by a page command. For example, the page (0x00) prior to a telemetry inquiry points to the internal μC data and pages (0x01) prior to a telemetry inquiry points to the BCM connected data. The BCM constantly polls it's data through the PMBus™.

The HV BCM enables the PMBus compatible host interface with an operating bus speed of up to 400 kHz. The HV BCM follows the PMBus command structure and specification.

PMBus™ Interface

Refer to “PMBus Power System Management Protocol Specification Revision 1.2, Part I and II” for complete PMBus specifications details visit <http://pmbus.org>.

Device Address

The PMBus address (ADDR Pin) should be set to one of a predetermined 16 possible addresses shown in the table below using a resistor between ADDR pin and SGND pin.

The BCM accepts only a fixed and persistent address and does not support SMBus address resolution protocol. At initial power-up, the BCM internal µC will sample the address pin voltage, and will hold this address until device power is removed.

ID	Slave Address	HEX	Recommended Resistor R _{ADDR} (Ω)
1	1010 000b	50h	487
2	1010 001b	51h	1050
3	1010 010b	52h	1870
4	1010 011b	53h	2800
5	1010 100b	54h	3920
6	1010 101b	55h	5230
7	1010 110b	56h	6810
8	1010 111b	57h	8870
9	1011 000b	58h	11300
10	1011 001b	59h	14700
11	1011 010b	5Ah	19100
12	1011 011b	5Bh	25500
13	1011 100b	5Ch	35700
14	1011 101b	5Dh	53600
15	1011 110b	5Eh	97600
16	1011 111b	5Fh	316000

Reported DATA Formats

The BCM internal µC employs a direct data format where all reported internal µC measurements are in Volts, Amperes, Degrees Celsius, or Seconds. The host uses the following PMBus specification to interpret received values metric prefixes. Note that the Coefficients command is not supported:

$$X = \left(\frac{I}{m}\right) \cdot (Y \cdot 10^{-R} - b)$$

Where:

X, is a “real world” value in units (A, V, °C, s)

Y, is a two's complement integer received from the internal µC

m, b and R are two's complement integers defined as follows:

Command	Code	m	R	b
TON_DELAY	60h	1	3	0
READ_VIN	88h	1	1	0
READ_IIN	89h	1	3	0
READ_VOUT	8Bh	1	1	0
READ_IOUT	8Ch	1	2	0
READ_TEMPERATURE_1	8Dh	1	0	0
READ_POUT	96h	1	0	0
MFR_VIN_MIN	A0h	1	0	0
MFR_VIN_MAX	A1h	1	0	0
MFR_VOUT_MIN	A4h	1	0	0
MFR_VOUT_MAX	A5h	1	0	0
MFR_IOUT_MAX	A6h	1	0	0
MFR_POUT_MAX	A7h	1	0	0
READ_K_FACTOR	D1h	65536	0	0
READ_BCM_ROUT	D4h	1	2	0

^[1] Default READ Output Voltage returned when BCM unit is disabled = -300 V.

^[2] Default READ Temperature returned when BCM unit is disabled = -273°C.

No special formatting is required when lowering the supervisory limits and warnings.

Supported Command List

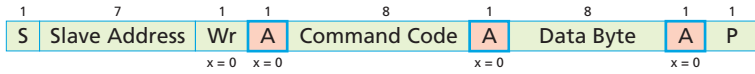
Command	Code	Function	Default Data Content	Data Bytes
PAGE	00h	Access BCM stored information for all connected devices	00h	1
OPERATION	01h	Turn BCMs on or off	80h	1
ON_OFF_CONFIG	02h	Defines startup when power is applied as well as immediate on/off control over the BCMs	1Dh	1
CLEAR_FAULTS	03h	Clear all BCM and all internal μ C faults	N/A	None
CAPABILITY	19h	Internal μ C PMBus™ key capabilities set by factory	20h	1
OT_FAULT_LIMIT	4Fh ^[1]	BCM over temperature protection	64h	2
OT_WARN_LIMIT	51h ^[1]	BCM over temperature warning	64h	2
VIN_OV_FAULT_LIMIT	55h ^[1]	BCM VIN overvoltage warning	64h	2
VIN_OV_WARN_LIMIT	57h ^[1]	BCM VIN overvoltage protection	64h	2
IIN_OC_FAULT_LIMIT	58h ^[1]	BCM IOUT overcurrent protection	64h	2
IIN_OC_WARN_LIMIT	5Dh ^[1]	BCM IOUT overcurrent warning	64h	2
TON_DELAY	60h ^[1]	Startup delay additional to any BCM fixed delays	00h	2
STATUS_BYTE	78h	Summary of BCM faults	00h	1
STATUS_WORD	79h	Summary of BCM fault conditions	00h	2
STATUS_IOUT	7Bh	BCM overcurrent fault status	00h	1
STATUS_INPUT	7Ch	BCM overvoltage and under voltage fault status	00h	1
STATUS_TEMPERATURE	7Dh	BCM over temperature and under temperature fault status	00h	1
STATUS_CML	7Eh	Internal μ C PMBus Communication fault	00h	1
STATUS_MFR_SPECIFIC	80h	Other BCM status indicator	00h	1
READ_VIN	88h	BCM input voltage	FFFFh	2
READ_IIN	89h	BCM input current	FFFFh	2
READ_VOUT	8Bh	BCM output voltage	FFFFh	2
READ_IOUT	8Ch	BCM output current	FFFFh	2
READ_TEMPERATURE_1	8Dh	BCM temperature	FFFFh	2
READ_POUT	96h	BCM output power	FFFFh	2
PMBUS_REVISION	98h	Internal μ C PMBus compatible revision	22h	1
MFR_ID	99h	Internal μ C ID	"VI"	2
MFR_MODEL	9Ah	Internal μ C or BCM model	Part Number	18
MFR_REVISION	9Bh	Internal μ C or BCM revision	FW and HW revision	18
MFR_LOCATION	9Ch	Internal μ C or BCM factory location	"AP"	2
MFR_DATE	9Dh	Internal μ C or BCM manufacturing date	"YYWW"	4
MFR_SERIAL	9Eh	Internal μ C or BCM serial number	Serial Number	16
MFR_VIN_MIN	A0h	BCM Minimum rated VIN	Varies per BCM	2
MFR_VIN_MAX	A1h	BCM Maximum rated VIN	Varies per BCM	2
MFR_VOUT_MIN	A4h	BCM Minimum rated VOUT	Varies per BCM	2
MFR_VOUT_MAX	A5h	BCM Maximum rated VOUT	Varies per BCM	2
MFR_IOUT_MAX	A6h	BCM Maximum rated IOUT	Varies per BCM	2
MFR_POUT_MAX	A7h	BCM Maximum rated POUT	Varies per BCM	2
BCM_EN_POLARITY	D0h ^[1]	Set BCM EN pin polarity	02h	1
READ_K_FACTOR	D1h	BCM K factor	Varies per BCM	2
READ_BCM_ROUT	D4h	BCM Rout	Varies per BCM	2
SET_ALL_THRESHOLDS	D5h ^[1]	Set BCM supervisory warning and protection thresholds	646464646464h	6
DISABLE_FAULT	D7h ^[1]	Disable BCM overvoltage, overcurrent or under voltage supervisory faults	00h	2

^[1] The BCM must be in a disabled state during a write message.

Command Structure Overview

Write Byte protocol:

The Host always initiates PMBus™ communication with a START bit. All messages are terminated by the Host with a STOP bit. In a write message, the master sends the slave device address followed by a write bit. Once the slave acknowledges, the master proceeds with the command code and then similarly the data byte.



- S** Start Condition
- Sr** Repeated start Condition
- Rd** Read
- Wr** Write
- X** Indicated that field is required to have the value of x
- A** Acknowledge (bit may be 0 for an ACK or 1 for a NACK)
- P** Stop Condition
- From Master to Slave
- From Slave to Master
- ... Continued next line

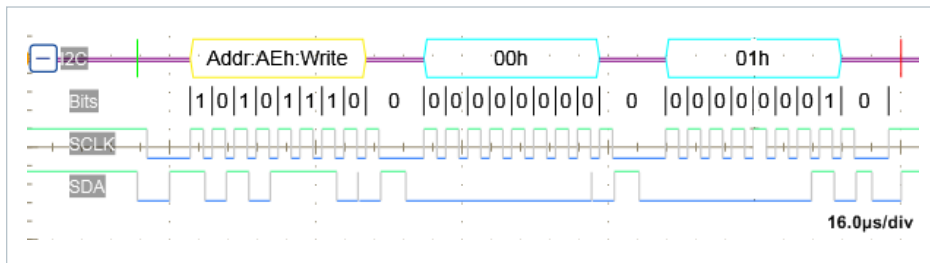


Figure 1 — PAGE COMMAND (00h), WRITE BYTE PROTOCOL

Read Byte protocol:

A Read message begins by first sending a Write Command, followed by a REPEATED START Bit and a slave Address. After receiving the READ bit, the internal µC begins transmission of the Data responding to the Command. Once the Host receives the requested Data, it terminates the message with a NACK preceding a stop condition signifying the end of a read transfer.

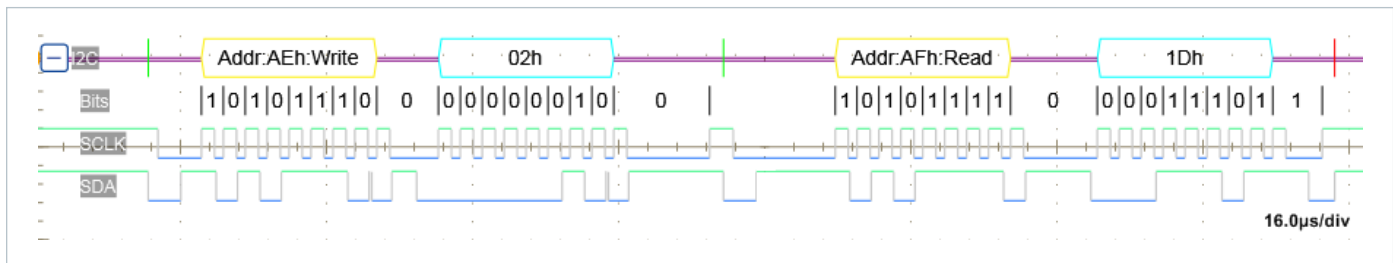
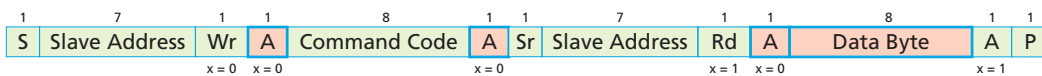


Figure 2 — ON_OFF_CONFIG COMMAND (02h), READ BYTE PROTOCOL

Write Word protocol:

When transmitting a word, the lowest order byte leads the highest order byte. Furthermore, when transmitting a Byte, the least significant bit (LSB) is sent last. Refer to System Management Bus (SMBus) specification version 2.0 for more details.

Note: Extended command and Packet Error Checking Protocols are not supported.

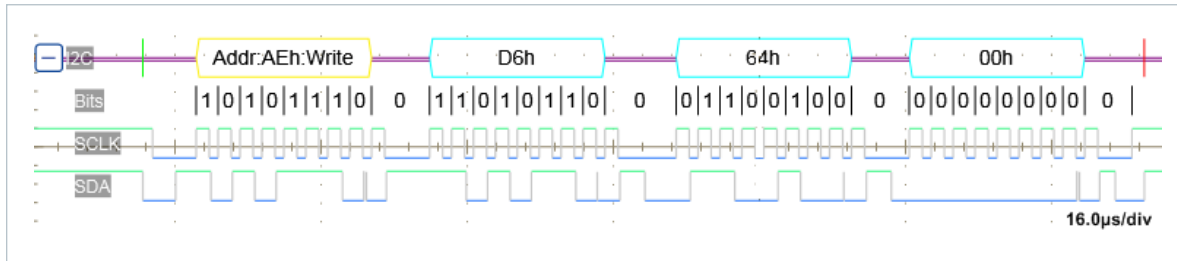
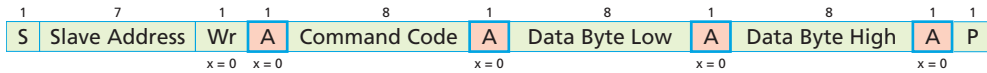


Figure 3 — TON_DELAY COMMAND (60h)_WRITE WORD PROTOCOL

Read Word protocol:

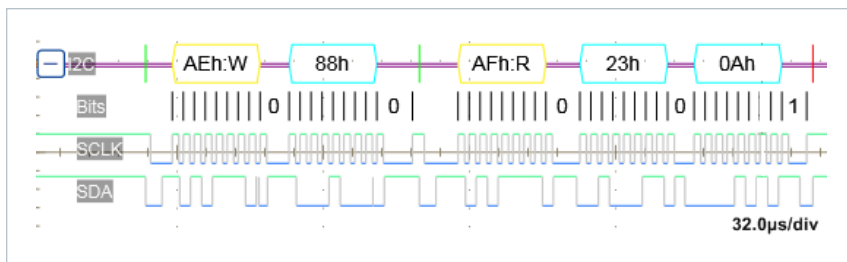
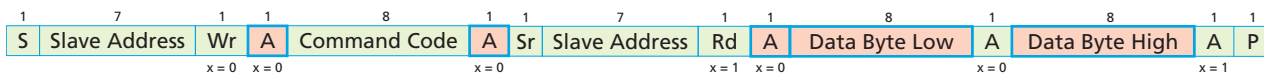


Figure 4 — MFR_VIN_MIN COMMAND (A0h)_READ WORD PROTOCOL

Write Block protocol:

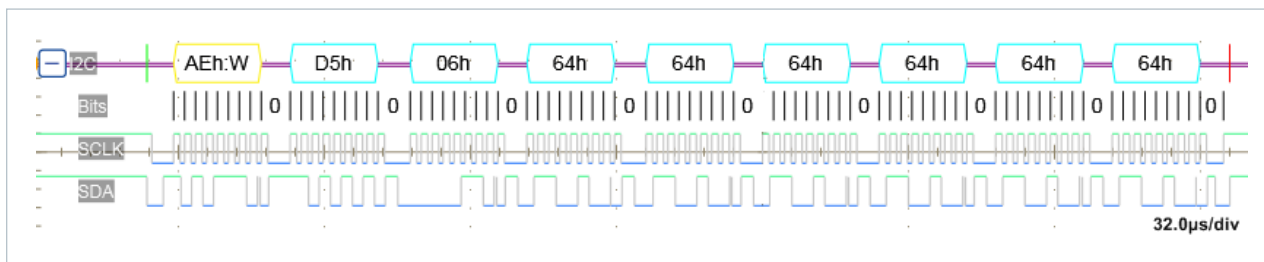
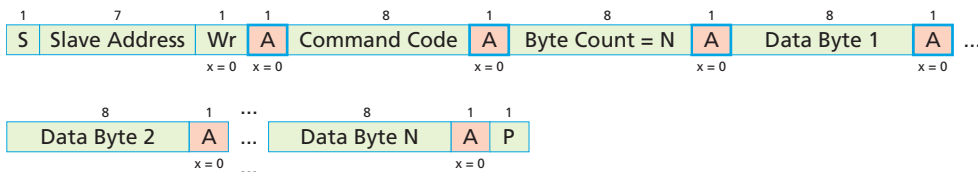


Figure 5 — SET_ALL_THRESHOLDS COMMAND (D5h)_WRITE BLOCK PROTOCOL

Read Block protocol:

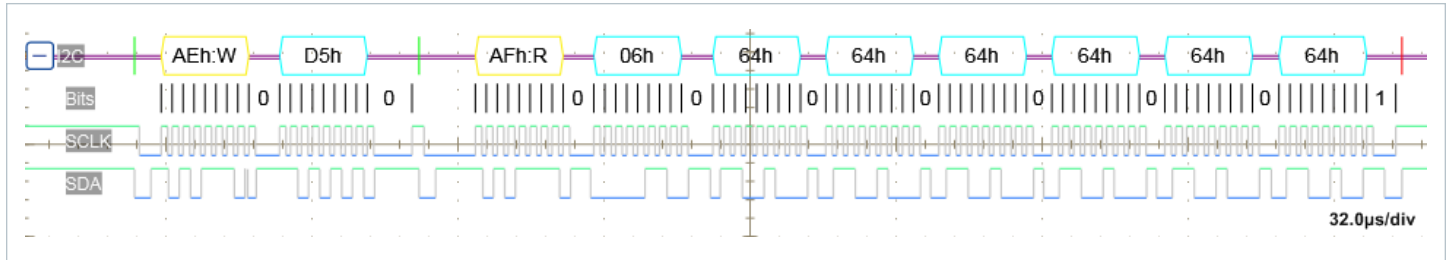
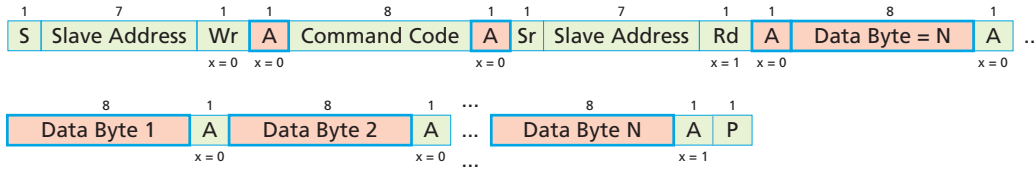


Figure 6 — SET_ALL_THRESHOLDS COMMAND (D5h)_READ BLOCK PROTOCOL

Write Group Command protocol:

Note that only one command per device is allowed in a group command.

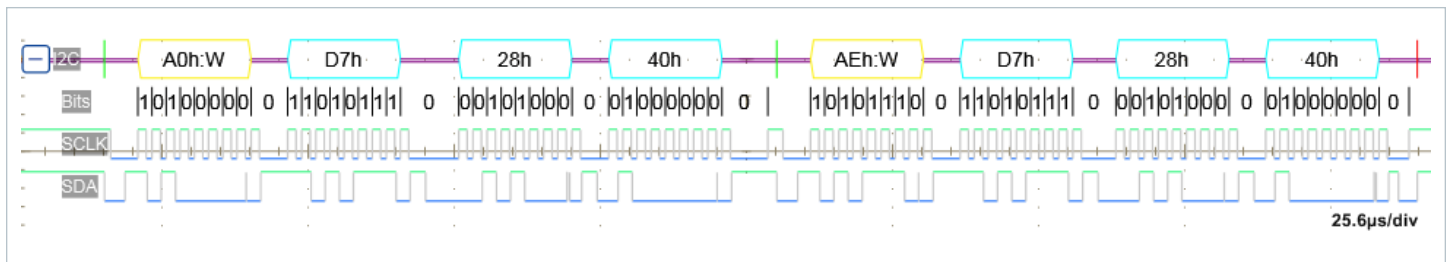
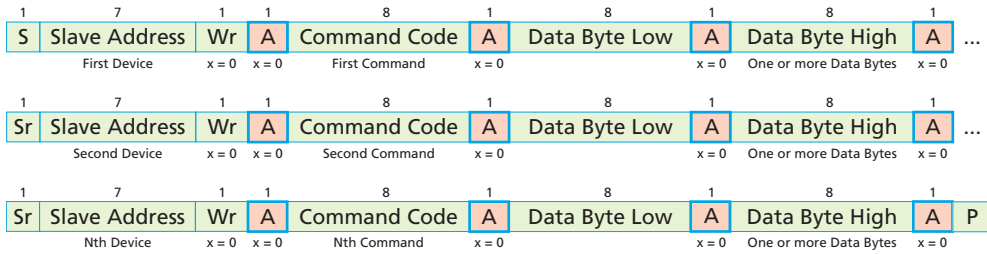


Figure 7 — DISABLE_FAULT COMMAND (D7h)_WRITE

Supported Commands Transaction type

A direct communication to the BCM internal μ C and a simulated communication to non-PMBus™ devices is enabled by a page command. Supported command access privileges with a pre-selected PAGE are defined in the following table. Deviation from this table generates a communication error in STATUS_CML register.

Command	Code	PAGE Data Byte Access Type	
		00h	01h
PAGE	00h	R/W	R/W
OPERATION	01h	R	R/W
ON_OFF_CONFIG	02h		R
CLEAR_FAULTS	03h	W	W
CAPABILITY	19h	R	
OT_FAULT_LIMIT	4Fh		R/W
OT_WARN_LIMIT	51h		R/W
VIN_OV_FAULT_LIMIT	55h		R/W
VIN_OV_WARN_LIMIT	57h		R/W
IIN_OC_FAULT_LIMIT	5Bh		R/W
IIN_OC_WARN_LIMIT	5Dh		R/W
TON_DELAY	60h		R/W
STATUS_BYTE	78h	R/W	R
STATUS_WORD	79h	R	R
STATUS_IOUT	7Bh	R	R/W
STATUS_INPUT	7Ch	R	R/W
STATUS_TEMPERATURE	7Dh	R	R/W
STATUS_CML	7Eh	R/W	
STATUS_MFR_SPECIFIC	80h	R	R/W
READ_VIN	88h		R
READ_IIN	89h	R	R
READ_VOUT	8Bh		R
READ_IOUT	8Ch	R	R
READ_TEMPERATURE_1	8Dh	R	R
READ_POUT	96h	R	R
PMBUS_REVISION	98h	R	
MFR_ID	99h	R	
MFR_MODEL	9Ah	R	R
MFR_REVISION	9Bh	R	R
MFR_LOCATION	9Ch	R	R
MFR_DATE	9Dh	R	R
MFR_SERIAL	9Eh	R	R
MFR_VIN_MIN	A0h	R	R
MFR_VIN_MAX	A1h	R	R
MFR_VOUT_MIN	A4h	R	R
MFR_VOUT_MAX	A5h	R	R
MFR_IOUT_MAX	A6h	R	R
MFR_POUT_MAX	A7h	R	R
BCM_EN_POLARITY	D0h		R/W
READ_K_FACTOR	D1h		R
READ_BCM_ROUT	D4h		R
SET_ALL_THRESHOLDS	D5h		R/W
DISABLE_FAULT	D7h		R/W

Page Command (00h)

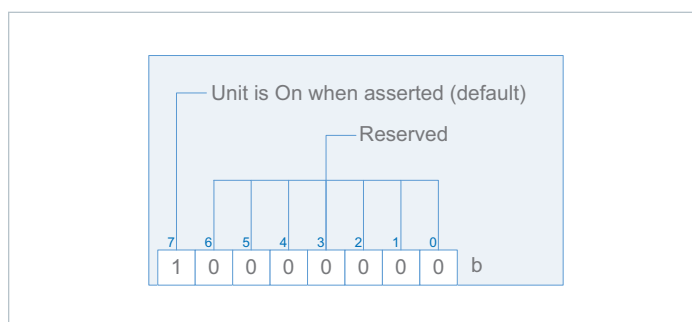
The page command data byte of 00h prior to a command call will address the internal μ C specific data and a page data byte of FFh would broadcast to all of the connected BCMs. The value of the Data Byte corresponds to the pin name trailing number with the exception of 00h and FFh.

Data Byte	Description
00h	μ C
01h	BCM

OPERATION Command (01h)

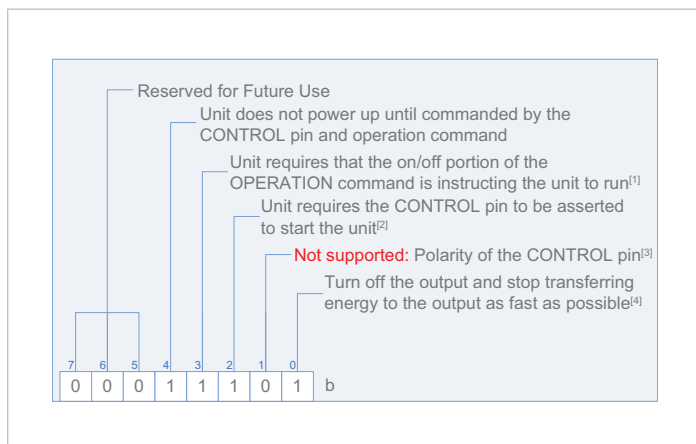
The Operation command can be used to turn on and off the connected BCM. Note that the host OPERATION command will not enable the BCM if the BCM EN pin is disabled in hardware with respect to the pre-set pin polarity. Only with the EN pin active, will the OPERATION command provide ON/OFF control.

If synchronous startup is required in the system, it is recommended to use the command from host PMBus in order to achieve simultaneous array startup.



This command accepts only two data values: 00h and 80h. If any other value is sent the command will be rejected and a CML Data error will result.

ON_OFF_CONFIG Command (02h)



[1] The BCM Enable pin is ALWAYS to be asserted for powerup. The BCM_EN_POLARITY command (D0h) bit[1] defines the logic level required for the control pin (i.e BCM Enable pin) to be asserted.

[2] With respect to the BCM EN Control Pin if used in system

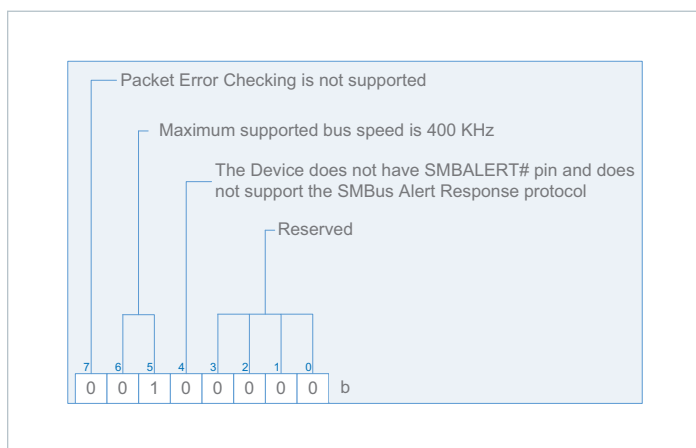
[3] See MFR_SPECIFIC_00 / BCM_EN_POLARITY to change the Polarity of the BCM Enable Pin

[4] The BCM powertrain once disabled cannot sink current

CLEAR_FAULTS Command (03h)

This command clears all status bits that have been previously set. Persistent or active faults are re-asserted again once cleared. All faults are latched once asserted in the internal µC. Registered faults will not be cleared when shutting down the BCM powertrain by recycling the BCM input voltage, or toggling the BCM EN pin, or sending the OPERATION command.

CAPABILITY Command (19h)



The internal µC returns a default value of 20h. This value indicates that the PMBus™ frequency supported is up to 400 KHz and that both Packet Error Checking (PEC) and SMBALERT# are not supported.

OT_FAULT_LIMIT Command (4Fh), OT_WARN_LIMIT Command (51h), VIN_OV_FAULT_LIMIT Command (55h), VIN_OV_WARN_LIMIT Command (57h), IIN_OC_FAULT_LIMIT Command (5Bh), IIN_OC_WARN_LIMIT Command (5Dh)

The values of these registers are set in non-volatile memory and can only be written when the BCMs are disabled.

The values of the above mentioned fault and warning are set by default to a 100% of the respective BCM model supervisory limits. However these limits can be set to a lower value. For example: In order for a limit percentage to be set to 80% one would send a write command with a (50h) Data Word.

Any values outside the range of (00h – 64h) sent by a host will be rejected, will not override the currently stored value and will set the Unsupported Data bit in STATUS_CML.

The SET_ALL_THRESHOLDS COMMAND (D5h) combines in one block over temperature fault and warning limits, V_{IN} overvoltage fault and warning limits as well as I_{OUT} overcurrent fault and warning limits. A delay prior to a read command of up to 200 ms following a write of new value is required.

The VIN_UV_WARN_LIMIT (58h) and VIN_UV_FAULT_LIMIT (59h) are set by the factory and cannot be changed by the host. However, a host can disable the under voltage setting using the DISABLE_FAULT COMMAND (D7h).

All FAULT_RESPONSE commands are unsupported. The BCM powertrain supervisory limits and powertrain protection will behave as described in the BCM datasheet. In general, once a fault is detected, the BCM powertrain will shut down and attempt to auto-restart after a predetermined delay.

TON_DELAY Command (60h)

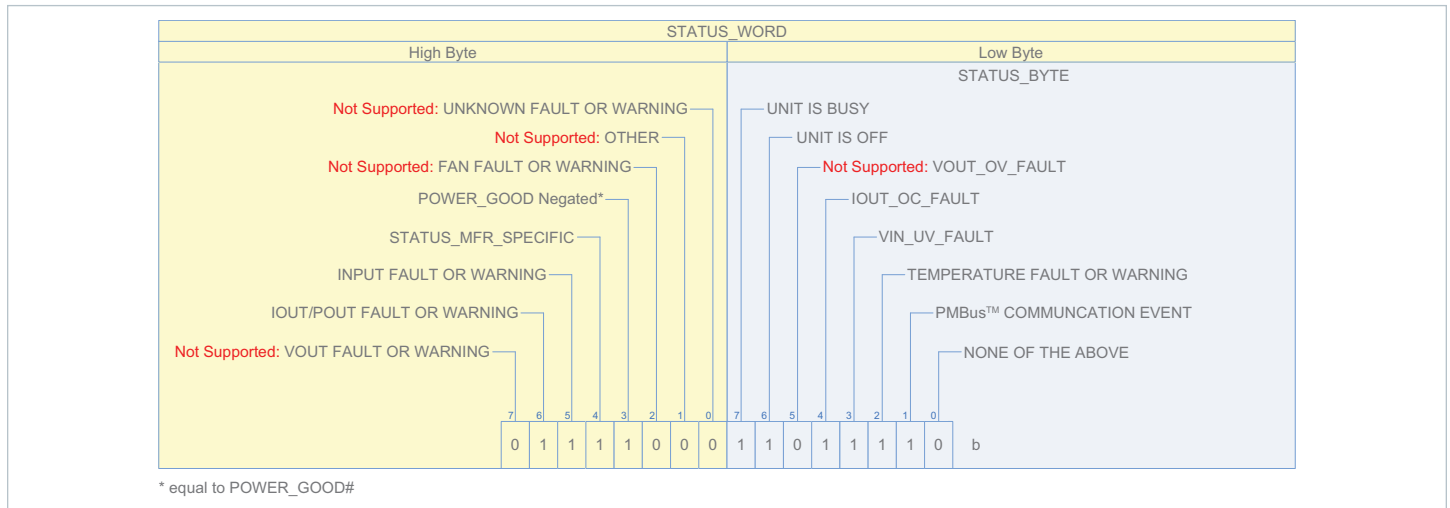
The value of this register word is set in non-volatile memory and can only be written when the BCMs are disabled.

The maximum possible delay is 100ms. Default value is set to (00h). Follow this equation below to interpret the reported value.

$$TON_DELAY_{ACTUAL} = t_{REPORTED} \cdot 10^{-3}(s)$$

Staggering startup in an array is possible with TON_DELAY Command. This delay will be in addition to any startup delay inherent in the BCM module. For example: startup delay from application of V_{IN} is typically 20 ms whereas startup with EN pin is typically 250 us. When TON_DELAY is greater than zero, the set delay will be added to both.

STATUS_BYTE (78h) and STATUS_WORD (79h)



All fault or warning flags, if set, will remain asserted until cleared by the host or once the internal μ C power is removed. This includes under voltage fault, overvoltage fault, overvoltage warning, overcurrent warning, over temperature fault, over temperature warning, under temperature fault, reverse operation, communication faults and analog controller shutdown fault.

Asserted status bits in all status registers, with the exception of STATUS_WORD and STATUS_BYTE, can be individually cleared. This is done by sending a data byte with one in the bit position corresponding to the intended warning or fault to be cleared. Refer to the PMBus™ Power System Management Protocol Specification – Part II – Revision 1.2 for details.

The POWER_GOOD# bit reflects the state of the device and does not reflect the state of the POWER_GOOD# signal limits. The POWER_GOOD_ON COMMAND (5Eh) and POWER_GOOD_OFF COMMAND (5Fh) are not supported. The POWER_GOOD# bit is set anytime the BCM is not in the enabled state, to indicate that the powertrain is inactive and not switching. The POWER_GOOD# bit is cleared when the BCM completes the enabling state, 5 ms after the powertrain is activated allowing for soft-start to elapse. POWER_GOOD# and OFF bits cannot be cleared as they always reflect the current state of the device.

When Page (00h) is used the POWER_GOOD# bit reflects the OR-ing of all active BCMs' POWER_GOOD# bits. When Page (01h – 04h) is used POWER_GOOD# is clear only when the BCM is active.

When Page (00h) is used UNIT IS OFF is SET when all BCMs are not active. When Page (01h – 04h) is used UNIT IS OFF is clear only when the BCM is active.

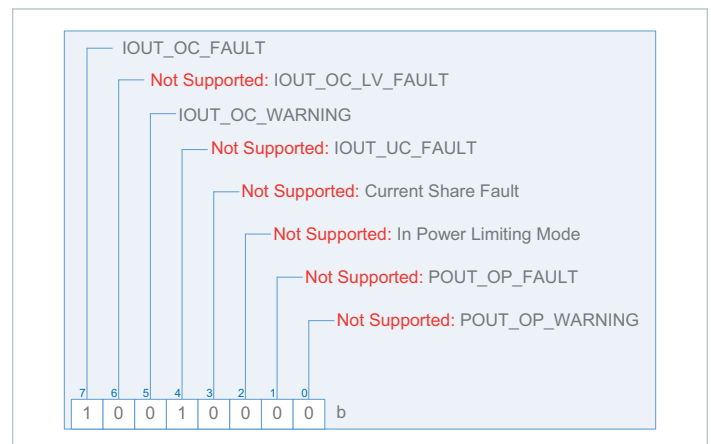
The Busy bit can be cleared using CLEAR_ALL Command (03h) or by writing either data value (40h, 80h) to PAGE (00h) using the STATUS_BYTE (78h).

Fault reporting, such as SMBALERT# signal output, and host notification by temporarily acquiring bus master status is not supported.

If the internal μ C is still powered, it will retain the last status it received from the BCM and this information will be available to the user via a PMBus Status request. This is in agreement with the PMBus standard which requires that status bits remain set until specifically cleared. Note that in this case where the BCM V_{IN} is lost, the status will always indicate an under voltage fault, in addition to any other fault that occurred.

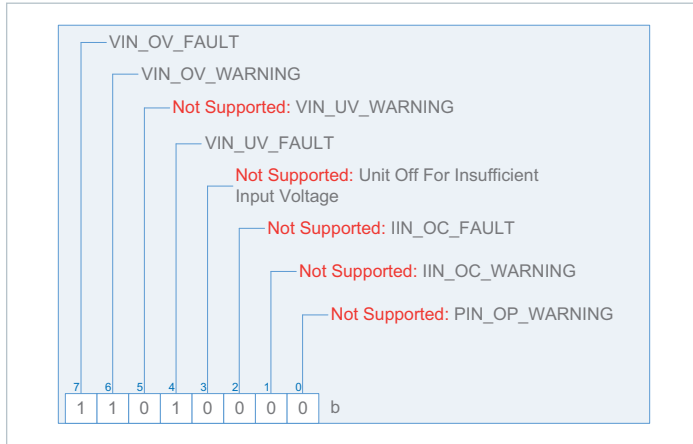
NONE OF THE ABOVE bit will be asserted if either the STATUS_MFR_SPECIFIC (80h) or the High Byte of the STATUS WORD is set.

STATUS_IOUT (7Bh)



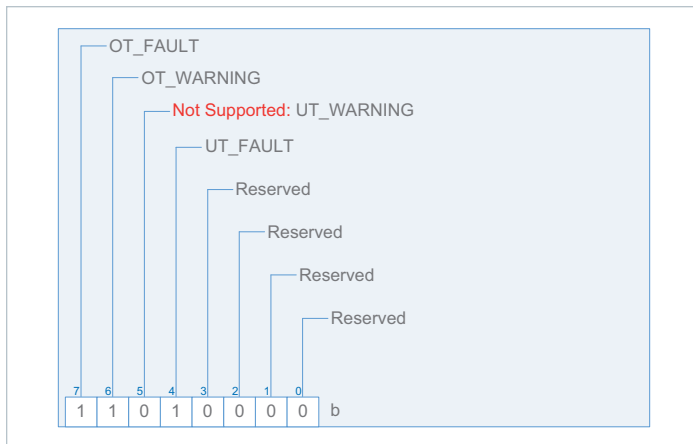
Unsupported bits are indicated above. A one indicates a fault.

STATUS_INPUT (7Ch)



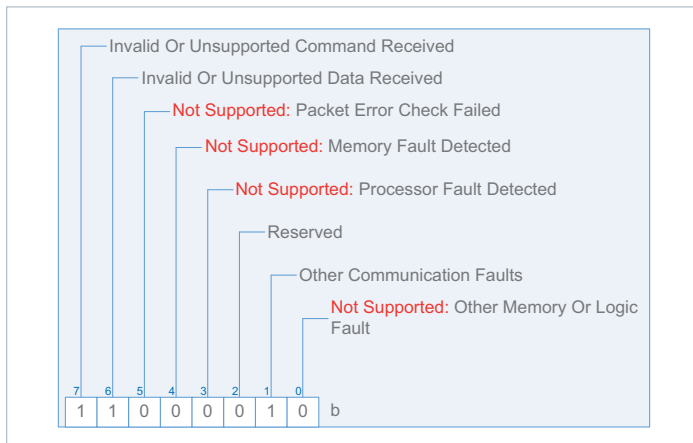
Unsupported bits are indicated above. A one indicates a fault.

STATUS_TEMPERATURE (7Dh)



Unsupported bits are indicated above. A one indicates a fault.

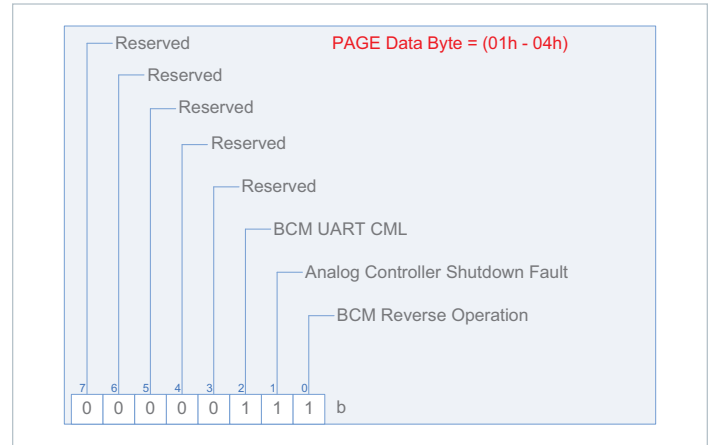
STATUS_CML (7Eh)



Unsupported bits are indicated above. A one indicates a fault.

The STATUS_CML data byte will be asserted when an unsupported PMBus™ command or data or other communication fault occurred.

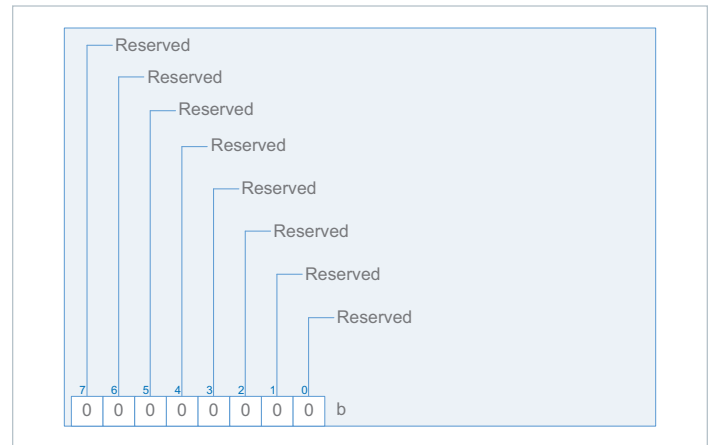
STATUS_MFR_SPECIFIC (80h)



The reverse operation bit, if asserted, indicates that the BCM is processing current in reverse. Reverse current reported value is not supported.

The BCM has analog protections and internal μC protections. The analog controller provides an additional layer of protection and has the fastest response time. The analog controller shutdown fault, when asserted, indicates that at least one of the powertrain protection faults is triggered. This fault will also be asserted if a disabled fault event occurs after asserting any bit using the DISABLE_FAULTS COMMAND.

The BCM UART is designed to operate with the internal μC UART. If the BCM UART CML is asserted, it may indicate a hardware or connection issue between both devices.



When PAGE COMMAND (00h) data byte is equal to (00h), the BCM Reverse operation, Analog Controller Shutdown Fault, and BCM UART CML bit will return OR-ing result of active BCMs. The BCM UART CML will also be asserted if any of the active BCMs stops responding. The BCM must communicate at least once to the internal μC in order to trigger this FAULT. The BCM UART CML can be cleared from the culprit BCM once the internal μC is able to communicate with it once again or can be cleared using PAGE (00h) CLEAR_FAULTS (03h) Command.

READ_VIN Command (88h)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's input voltage in the following format:

$$V_{VIN_ACTUAL} = V_{VIN_REPORTED} \cdot 10^{-1}(V)$$

READ_IIN Command (89h)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's input current in the following format:

$$I_{IN_ACTUAL} = I_{IN_REPORTED} \cdot 10^{-3}(A)$$

If PAGE data byte is equal (00h) command will return the sum of active BCMs' input current.

READ_VOUT Command (8Bh)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's output voltage in the following format:

$$V_{VOUT_ACTUAL} = V_{VOUT_REPORTED} \cdot 10^{-1}(V)$$

READ_IOUT Command (8Ch)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's output current in the following format:

$$I_{IOUT_ACTUAL} = I_{IOUT_REPORTED} \cdot 10^{-2}(A)$$

If PAGE data byte is equal (00h) command will return the sum of active BCMs' output current.

READ_TEMPERATURE_1 Command (8Dh)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's temperature in the following format:

$$T_{ACTUAL} = \pm T_{REPORTED} (^{\circ}C)$$

If PAGE data byte is equal (00h) command will return the maximum temperature of active BCMs.

READ_POUT Command (96h)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's output power in the following format:

$$POUT_{ACTUAL} = POUT_{REPORTED} (W)$$

If PAGE data byte is equal to (00h) command will return the sum of active BCMs' output power.

MFR_VIN_MIN Command (A0h), MFR_VIN_MAX Command (A1h), MFR_VOUT_MIN Command (A4h), MFR_VOUT_MAX Command (A5h), MFR_IOUT_MAX Command (A6h), MFR_POUT_MAX Command (A7h)

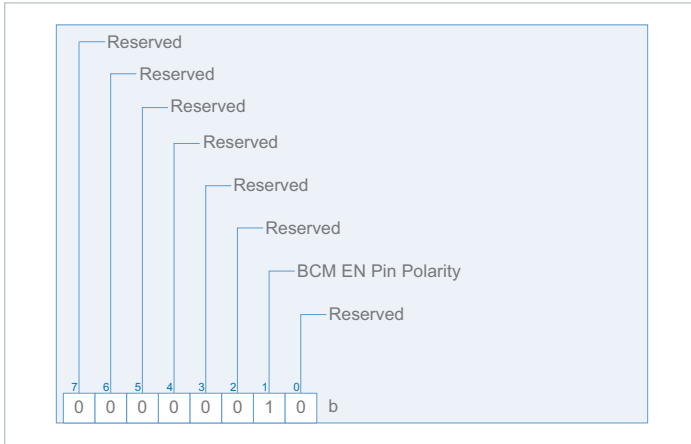
These values are set by the factory and indicate the device input output voltage and output current range and output power capacity.

The internal μC will report rated BCM input voltage minimum and maximum in Volts, output voltage minimum and maximum in Volts, output current maximum in Amperes and output power maximum in Watts.

If PAGE data byte is equal to (00h) then:

- MFR_VIN_MIN COMMAND (A0h) will return the highest MFR_VIN_MIN of all active BCMs
- MFR_VIN_MAX COMMAND (A1h) will return the lowest MFR_VIN_MAX of all active BCMs
- MFR_VOUT_MIN COMMAND (A4h) will return the highest MFR_VOUT_MIN of all active BCMs
- MFR_VOUT_MAX COMMAND (A5h) will return the lowest MFR_VOUT_MAX of all active BCMs
- MFR_IOUT_MAX COMMAND (A6h) will return the SUM of MFR_IOUT_MAX of all active BCMs
- MFR_POUT_MAX COMMAND (A7h) will return the SUM of MFR_POUT_MAX of all active BCMs

BCM_EN_POLARITY Command (D0h)



The value of this register is set in non-volatile memory and can only be written when the BCMs are disabled.

When PAGE COMMAND (00h) data byte is equal to (01h - 04h), this command defines the polarity of the EN pin. If BCM_EN_POLARITY is set, the BCM will startup once VIN is greater than the under voltage threshold.

The BCM EN PIN is internally pulled-up to 3.3V. If the BCM_EN_POLARITY is cleared, an external pull-down is then required. Applying VIN greater than the under voltage threshold will not suffice to start the BCM.

READ_K_FACTOR Command (D1h)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's K factor in the following format:

$$K_FACTOR_{ACTUAL} = K_FACTOR_{REPORTED} \cdot 2^{-16}(V/V)$$

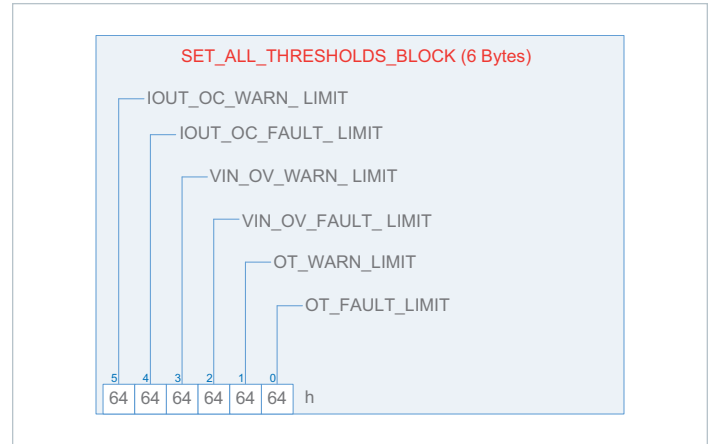
The K factor is defined in a BCM to represent the ratio of the transformer winding and hence is equal to V_{OUT} / V_{IN} .

READ_BCM_ROUT Command (D4h)

If PAGE data byte is equal to (01h - 04h) command will return a reported individual BCM's output resistance in the following format:

$$BCM_ROUT_{ACTUAL} = BCM_ROUT_{REPORTED} \cdot 10^{-5}(\Omega)$$

SET_ALL_THRESHOLDS Command (D5h)



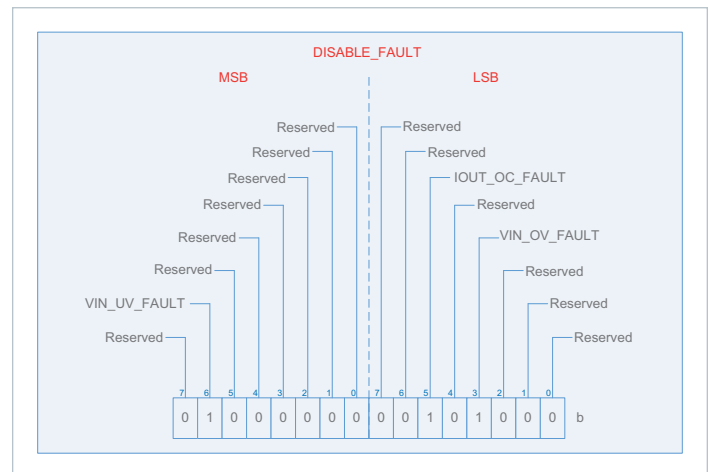
Values of this register block is set in non-volatile memory and can only be written when the BCMs are disabled.

This command provides a convenient way to configure all the limits, or any combination of limits described previously using one command.

Vin Overvoltage, Overcurrent and over-Temperature values are all set to 100% of the BCM datasheet supervisory limits by default and can only be set to a lower percentage.

To leave a particular threshold unchanged, set the corresponding threshold data byte to a value greater than (64h).

DISABLE_FAULT Command (D7h)



Unsupported bits are indicated above. A one indicates that the supervisory fault associated with the asserted bit is disabled.

The value of these registers is set in non-volatile memory and can only be written when the BCMs are disabled.

This command allows the host to disable the supervisory faults and respective statuses. It does not disable the powertrain analog protections or warnings with respect to the set limits in the SET_ALL_THRESHOLDS Command.

The input under-voltage can only be disabled to a pre-set low limit as shown in the functional reporting range in the BCM data sheet.

The internal μ C Implementation vs. PMBus™ Specification Rev 1.2

The internal μ C is an I²C compliant, SMBus™ compatible device and PMBus command compliant device. This section denotes some deviation, perceived as differences from the PMBus Part I and Part II specification Rev 1.2.

1. The internal μ C meets all Part I and II PMBus specification requirements with the following differences to the transport requirement.

Unmet DC parameter Implementation vs SMBus™ spec						
Symbol	Parameter	D44TL1A0		SMBus™ Rev 2.0		Units
		Min	Max	Min	Max	
V _{IL} ^[a]	Input Low Voltage	-	0.99	-	0.8	V
V _{IH} ^[a]	Input High Voltage	2.31	-	2.1	V _{VDD_IN}	V
I _{LEAK_PIN} ^[b]	Input Leakage per Pin	10	22	-	±5	μA

^[a] V_{VDD_IN} = 3.3 V

^[b] V_{BUS} = 5 V

2. The internal μ C accepts 38 PMBus command codes. Implemented commands execute functions as described in the PMBus specification.

- Deviations from the PMBus specification:
 - a. Section 15, fault related commands
 - The limits and Warnings unit implemented is percentage (%) a range from decimal (0-100) of the factory set limits.

3. The internal μ C unsupported PMBus command code response as described in the Fault Management and Reporting:

- Deviations from the PMBus specification:
 - a. PMBus section 10.2.5.3, exceptions
 - The busy bit of the STATUS_BYTE as implemented can be cleared (80h). In order to maintain compatibility with the specification (40h) can also be used.
 - Manufacturer Implementation of the PMBus Spec
 - a. PMBus section 10.5, setting the response to a detected fault condition
 - All powertrain responses are pre-set and cannot be changed. Refer to the BCM datasheet for details.
 - b. PMBus section 10.6, reporting faults and warnings to the Host
 - SMBALERT# signal and Direct PMBus Device to Host Communication are not supported. However, the Digital Supervisor will set the corresponding fault status bits and will wait for the host to poll.
 - c. PMBus section 10.7, clearing a shutdown due to a fault
 - There is no RESET pin or EN pin in the internal μ C. Cycling power to the internal μ C will not clear a BCM Shutdown. The BCM will clear itself once the fault condition is removed. Refer to the BCM datasheet for details.
 - d. PMBus Section 10.8.1, corrupted data transmission faults:
 - Packet error checking is not supported.

Data Transmission Faults Implementation

This section describes data transmission faults as implemented in the internal μ C.

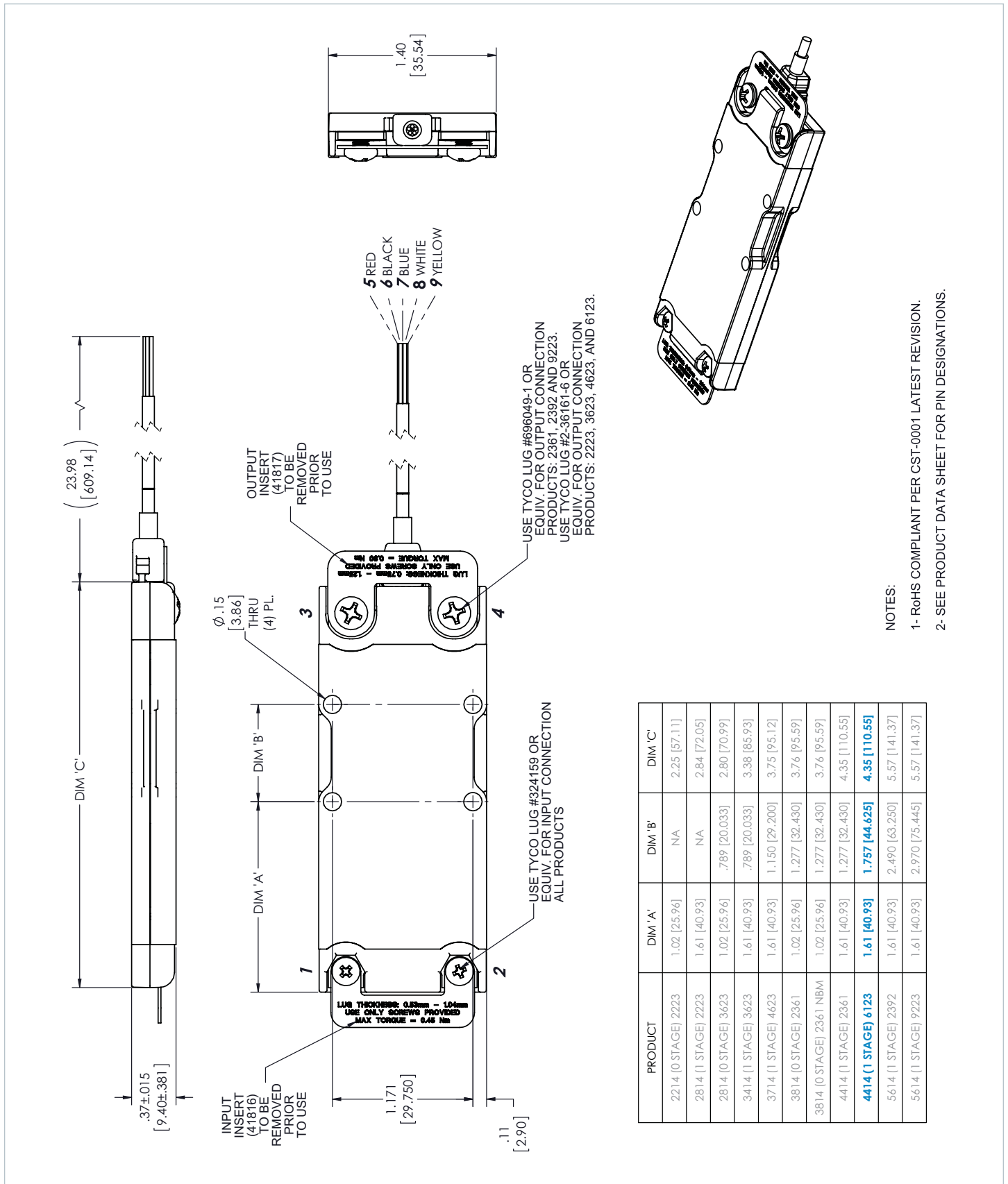
Section	Description	Response to Host		STATUS_BYTE	STATUS_CML		Notes
		NAK	FFh	CML	Other Fault	Unsupported Data	
10.8.1	Corrupted data						No response; PEC not supported
10.8.2	Sending too few bits			X	X		
10.8.3	Reading too few bits			X	X		
10.8.4	Host sends or reads too few bytes			X	X		
10.8.5	Host sends too many bytes	X		X		X	
10.8.6	Reading too many bytes		X	X	X		
10.8.7	Device busy	X	X				Device will ACK own address BUSY bit in STATUS_BYTE even if STATUS_WORD is set

Data Content Faults Implementation

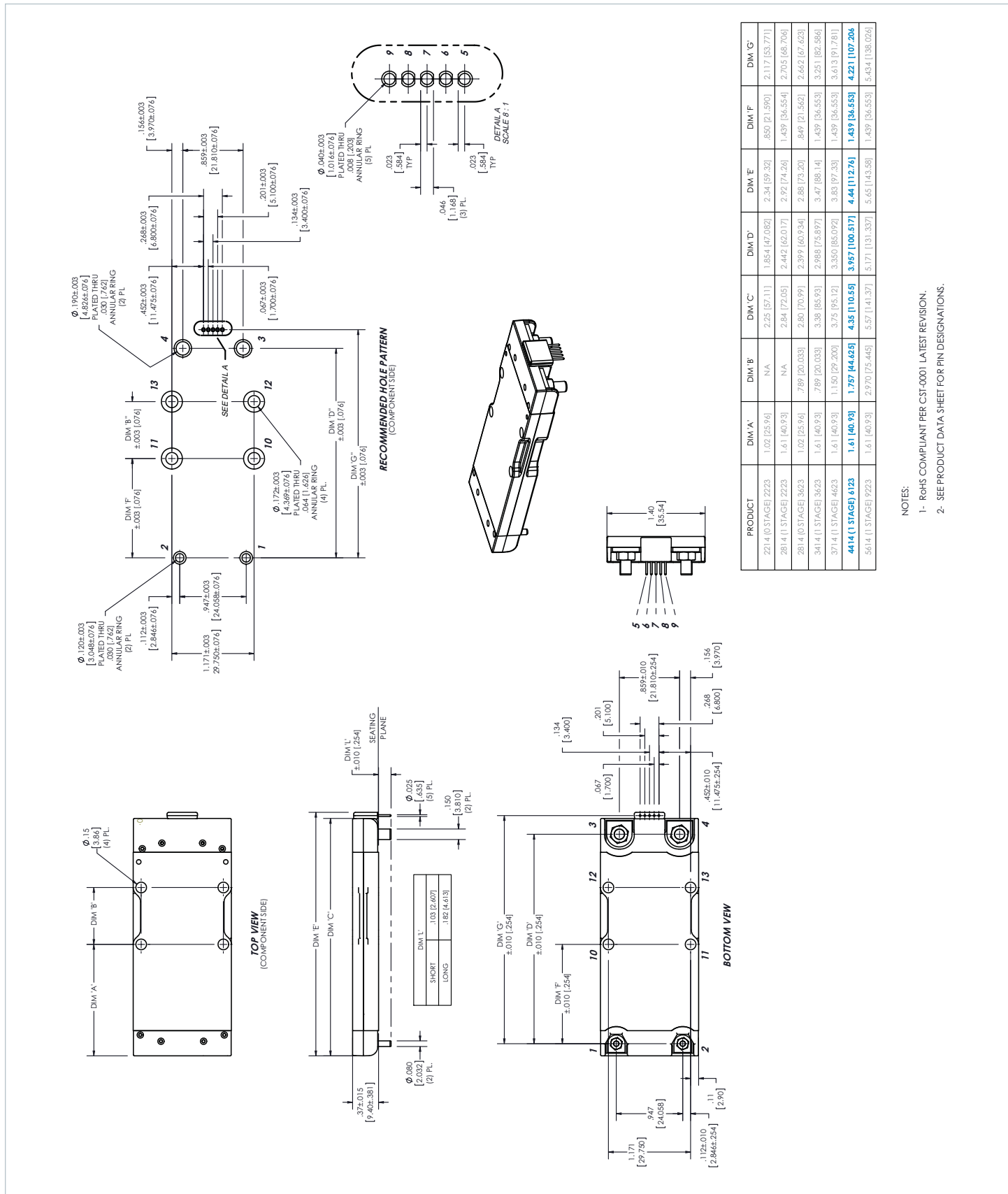
This section describes data content fault as implemented in the internal μ C.

Section	Description	Response to Host	STATUS_BYTE	STATUS_CML			Notes
		NAK	CML	Other Fault	Unsupported Command	Unsupported Data	
10.9.1	Improperly Set Read Bit In The Address Byte	X	X	X			
10.9.2	Unsupported Command Code	X	X		X		
10.9.3	Invalid or Unsupported Data		X			X	
10.9.4	Data Out of Range		X			X	
10.9.5	Reserved Bits						No response; not a fault

BCM in VIA Package Chassis (Lug) Mount Package Mechanical Drawing



BCM in VIA Package PCB (Board) Mount Package Mechanical Drawing and Recommended Hole Pattern



NOTES:
 1- ROHS COMPLIANT PER CST-0001 LATEST REVISION.
 2- SEE PRODUCT DATA SHEET FOR PIN DESIGNATIONS.

Revision History

Revision	Date	Description	Page Number(s)
1.0	03/3/16	Initial release	n/a

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